

5.28

a.

$$\frac{\sigma}{\omega \epsilon} = \frac{1}{2\pi(10^6)(\frac{81 \times 10^{-9}}{36\pi})} = \frac{18}{81} \times 10^3 = \frac{2}{9} \times 10^3 \gg 1 \Rightarrow \text{Good conductor}$$

$$\eta = \sqrt{\frac{\omega \mu}{2\sigma}} (1+j) = \sqrt{\frac{2\pi \times 10^6 (4\pi \times 10^{-7})}{2(1)}} (1+j) = 1.987(1+j) = 2.810 \angle 45^\circ$$

$$\Gamma = \frac{\eta - \eta_0}{\eta + \eta_0} = \frac{1.987(1+j) - 377}{1.987(1+j) + 377} = \frac{-375.013 + j1.987}{378.987 + j1.987} = 0.9895 \angle 179.4^\circ$$

$$E^r = |\Gamma| E^i = 0.9895 (1 \times 10^{-3}) = 0.9895 \times 10^{-3} \text{ V/m}$$

b.

$$SWR = \frac{1+|\Gamma|}{1-|\Gamma|} = \frac{1+0.9895}{1-0.9895} = 189.476$$

c.

$$S^i = \frac{1}{2\eta} |E^i|^2 = \frac{1}{2(377)} (10^{-3})^2 = 1.326 \times 10^{-9} \text{ W/m}^2$$

$$S^r = \frac{1}{2\eta_0} |E^r|^2 = \frac{1}{2\eta} |\Gamma|^2 |E^i|^2 = |\Gamma|^2 S^i = (0.9895)^2 (1.326 \times 10^{-9}) = 1.2983 \times 10^{-9} \text{ W/m}^2$$

$$d. \quad T = \frac{2\eta}{\eta + \eta_0} = \frac{2(2.810 \angle 45^\circ)}{378.987 + j1.987} = \frac{2(2.810 \angle 45^\circ)}{378.992 \angle 0.3^\circ} = 0.0148 \angle 44.7^\circ$$

$$E^t = |T| E^i = 0.0148 \times 10^{-3} \text{ V/m}$$

$$e. \quad S_L^t = S^i - S^r = S^i - |\Gamma|^2 S^i = (1 - |\Gamma|^2) S^i = (1 - 0.9895^2) 1.326 \times 10^{-9} = 0.0277 \times 10^{-9} \text{ W/m}^2$$

$$f. \quad \delta = \sqrt{\frac{2}{\omega \mu \sigma}} = \sqrt{\frac{2}{\pi f \mu \sigma}} = \sqrt{\frac{2}{\pi (10^6) (4\pi \times 10^{-7}) (1)}} = \frac{1}{2\pi} \sqrt{10} = 0.5033 \text{ m}$$

$$g. \quad \lambda = 2\pi \sqrt{\frac{2}{\omega \mu \sigma}} = 2\pi \delta = 2\pi (0.5033) = 3.1623 \text{ m}$$

$$d = 20\lambda = 2(3.1623) = 63.2455 \text{ m}$$

$$h. \quad v = \sqrt{\frac{2\omega}{\mu \sigma}} = \sqrt{\frac{2(2\pi \times 10^6)}{4\pi \times 10^{-7} (1)}} = \sqrt{10} (10^6) = 3.1623 \times 10^6 \text{ m/sec.}$$

$$t = \frac{d}{v} = \frac{100}{3.1623 \times 10^6} = 31.623 \times 10^{-6} \text{ sec.}$$

$$i. \quad \frac{v}{v_0} = \frac{3.1623 \times 10^6}{3 \times 10^8} = 1.054 \times 10^{-2}$$



5.30

a. For right hand circular $\psi = 90^\circ$

$$b. \underline{H}^i = (j\hat{a}_y - \hat{a}_z) \frac{E_0}{\eta_0} e^{+j\beta_0 x}$$

$$c. \frac{\sigma}{\omega \epsilon} = \frac{10^{-1}}{2\pi(10^9) \left(\frac{91 \times 10^{-12}}{36\pi} \right)} = \frac{18}{81} \times 10^{-1} \ll 1 \Rightarrow \text{Good dielectric}$$

$$\Gamma = \frac{\eta - \eta_0}{\eta + \eta_0} = \frac{\sqrt{\mu_0/\epsilon} - \sqrt{\mu_0/\epsilon_0}}{\sqrt{\mu_0/\epsilon} + \sqrt{\mu_0/\epsilon_0}} = \frac{\sqrt{\epsilon_0} - \sqrt{\epsilon}}{\sqrt{\epsilon_0} + \sqrt{\epsilon}} = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}} = \frac{1 - \sqrt{81}}{1 + \sqrt{81}} = \frac{1 - 9}{1 + 9} = -0.8$$

$$\underline{E}^r = -0.8(\hat{a}_y + j\hat{a}_z) E_0 e^{-j\beta_0 x}$$

$$\underline{H}^r = -\frac{0.8}{377}(\hat{a}_z - j\hat{a}_y) E_0 e^{-j\beta_0 x} = 2.122 \times 10^{-3} (j\hat{a}_y - \hat{a}_z) E_0 e^{-j\beta_0 x}$$

cont'd.

d. Circular, CCW

$$e. T = \frac{2\eta}{\eta + \eta_0} = \frac{2\sqrt{\mu_0/\epsilon}}{\sqrt{\mu_0/\epsilon} + \sqrt{\mu_0/\epsilon_0}} = 2 \frac{\frac{1}{\sqrt{\epsilon}}}{\frac{1}{\sqrt{\epsilon}} + \frac{1}{\sqrt{\epsilon_0}}} = 2 \frac{\sqrt{\epsilon_0}}{\sqrt{\epsilon_0} + \sqrt{\epsilon}} = \frac{2}{1 + \sqrt{\epsilon_r}} = \frac{2}{1 + 9} = 0.2$$

$$\underline{E}^t = 0.2(\hat{a}_y + j\hat{a}_z) e^{-\gamma x} = 0.2(\hat{a}_y + j\hat{a}_z) e^{-(\alpha + j\beta)x}$$

$$\underline{H}^t = \frac{0.2}{377} (j\hat{a}_y - \hat{a}_z) E_0 e^{-(\alpha + j\beta)x} = 5.3 \times 10^{-4} (j\hat{a}_y - \hat{a}_z) E_0 e^{-(\alpha + j\beta)x}$$

$$\alpha \approx \frac{\sigma}{2} \sqrt{\frac{\mu_0}{\epsilon}} = \frac{10^{-1}}{2} \left(\frac{377}{9} \right) = 2.0944 \text{ N/m}$$

$$\beta \approx \omega \sqrt{\mu_0 \epsilon} = \beta_0 \sqrt{81} = 9\beta_0 = 9\omega \sqrt{\mu_0 \epsilon_0} = \frac{9(2\pi \times 10^9)}{3 \times 10^8} = 9(20.944) = 188.5 \text{ rad}$$

f. Circular, CW

$$g. \frac{S^r}{S^i} = |\Gamma|^2 = |-0.8|^2 = 0.64 = 64\%$$

$$\frac{S^t}{S^i} = 1 - |\Gamma|^2 = (1 - |-0.8|^2) = 0.36 = 36\%$$

CHAPTER 1

7.1 a. $H_\phi = \frac{E_\theta}{\eta} = j \frac{\beta I_0 l e^{-j\beta r}}{4\pi r} \sin\theta [2\cos(\beta h \cos\theta)]$

b. $S_{av} = \frac{1}{2} \text{Re}[E \times H^*] = \frac{1}{2} \text{Re}[\hat{a}_\theta E_\theta \times \hat{a}_\phi H_\phi^*] = \frac{\hat{a}_r}{2} \text{Re}[E_\theta H_\phi^*]$
 $= \hat{a}_r \frac{1}{2} \text{Re}[E_\theta \frac{E_\theta^*}{\eta}] = \hat{a}_r \frac{1}{2} \frac{|E_\theta|^2}{\eta}$

$S_{av} = \hat{a}_r \frac{\eta}{2} \left| \frac{\beta I_0 l}{4\pi r} \right|^2 \sin^2\theta [2\cos(\beta h \cos\theta)]^2 = \hat{a}_r 2\eta \left| \frac{\beta I_0 l}{4\pi r} \right|^2 \sin^2\theta \cos^2(\beta h \cos\theta)$

c. $P_{av} = \iiint_S S_{av} \cdot d\mathbf{s} = \int_0^{2\pi} \int_0^{\pi/2} \hat{a}_r S_{av} \cdot \hat{a}_r r^2 \sin\theta d\theta d\phi = \int_0^{2\pi} \int_0^{\pi/2} S_{av} r^2 \sin\theta d\theta d\phi$

$P_{av} = 4\pi \eta \left| \frac{\beta I_0 l}{4\pi} \right|^2 \int_0^{\pi/2} \sin^3\theta \cos^2(\beta h \cos\theta) d\theta = \pi \eta \left| \frac{I_0 l}{\lambda} \right|^2 \int_0^{\pi/2} \sin^3\theta \cos^2(\beta h \cos\theta) d\theta$

$P_{rad} = P_{av} = \pi \eta \left| \frac{I_0 l}{\lambda} \right|^2 I$

where $I = I_1 + I_2 = \int_0^{\pi/2} \sin^3\theta \cos^2(\beta h \cos\theta) d\theta = \int_0^{\pi/2} \sin^3\theta \left[\frac{1 + \cos(2\beta h \cos\theta)}{2} \right] d\theta$

$I = I_1 + I_2 = \frac{1}{2} \int_0^{\pi/2} \sin^3\theta d\theta + \frac{1}{2} \int_0^{\pi/2} \sin^3\theta \cos(2\beta h \cos\theta) d\theta$

$I_1 = \frac{1}{2} \int_0^{\pi/2} \sin^3\theta d\theta = -\frac{1}{6} \cos\theta (\sin^2\theta + 2) \Big|_0^{\pi/2} = \frac{1}{3}$

$I_2 = \frac{1}{2} \int_0^{\pi/2} \sin^3\theta \cos(2\beta h \cos\theta) d\theta = \frac{1}{2} \int_0^{\pi/2} \sin^2\theta \cos(2\beta h \cos\theta) \sin\theta d\theta$

Let $u = \sin^2\theta$ $du = 2\sin\theta \cos\theta d\theta$ $dv = -\frac{\cos(2\beta h \cos\theta)}{2\beta h} d(2\beta h \cos\theta)$, $v = -\frac{1}{2\beta h} \sin(2\beta h \cos\theta)$

Thus

$I_2 = -\underbrace{\frac{\sin^2\theta}{2\beta h} \sin(2\beta h \cos\theta)}_{=0} \Big|_0^{\pi/2} + \frac{1}{2\beta h} \int_0^{\pi/2} \cos\theta \sin(2\beta h \cos\theta) \sin\theta d\theta$

Let $u = \cos\theta$ $du = -\sin\theta d\theta$ $dv = -\frac{1}{2\beta h} \sin(2\beta h \cos\theta) d(2\beta h \cos\theta)$
 $v = \frac{1}{2\beta h} \cos(2\beta h \cos\theta)$

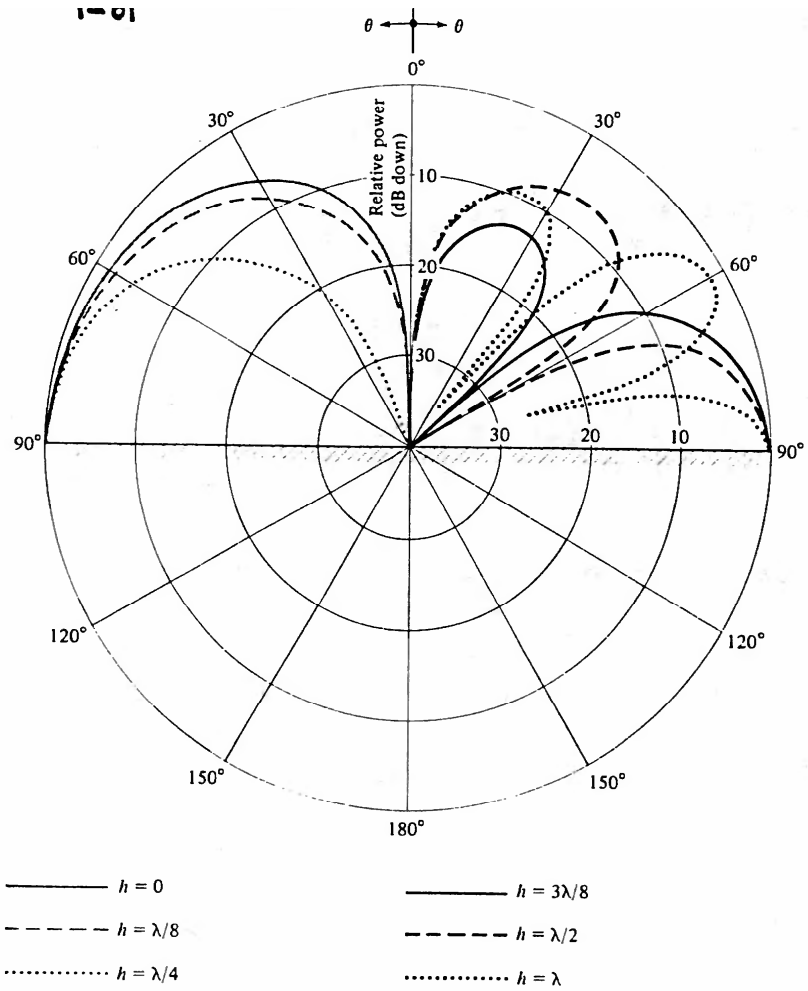
$I_2 = 0 + \frac{1}{2\beta h} \left\{ \frac{\cos\theta}{2\beta h} \cos(2\beta h \cos\theta) \Big|_0^{\pi/2} + \frac{1}{2\beta h} \int_0^{\pi/2} \cos(2\beta h \cos\theta) \sin\theta d\theta \right\}$

$I_2 = \frac{1}{2\beta h} \left\{ -\frac{1}{2\beta h} \cos(2\beta h) - \frac{1}{(2\beta h)^2} \sin(2\beta h \cos\theta) \Big|_0^{\pi/2} \right\} = \left\{ -\frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right\}$

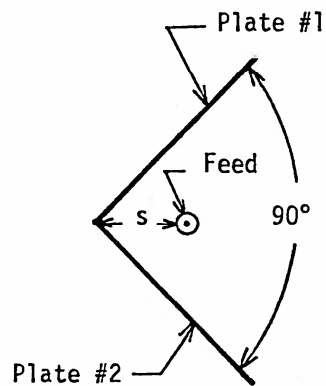
Therefore

$P_{rad} = \pi \eta \left| \frac{I_0 l}{\lambda} \right|^2 \left[\frac{1}{3} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right]$

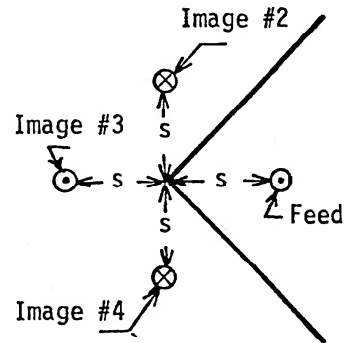
7.3



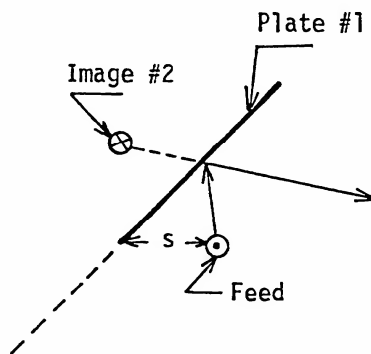
- 7.5** a. The number of images (3 of them), their polarizations, and their positions are shown below.



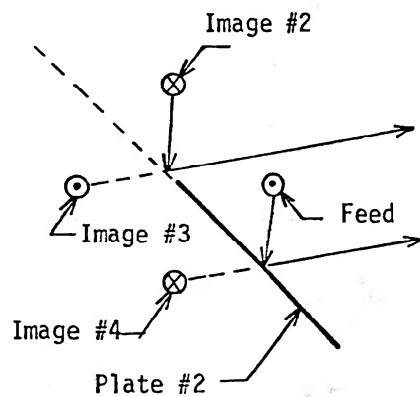
(a) 90° corner reflector



(b) images for 90° corner reflector



(c) placement of image #2
(due to feed)



(d) placement of images #4
and #3 (due to feed and
image #2)

7.5 cont'd. b. The total field is derived by summing the contributions from the feed and its images. Thus

$$E^t(r, \theta, \phi) = E_1(r_1, \theta, \phi) + E_2(r_2, \theta, \phi) + E_3(r_3, \theta, \phi) + E_4(r_4, \theta, \phi) \quad (1)$$

In the far zone, the normalized field can be written as

$$E^t(r, \theta, \phi) = f(\theta, \phi) \frac{e^{-j\beta r_1}}{r_1} - f(\theta, \phi) \frac{e^{-j\beta r_2}}{r_2} + f(\theta, \phi) \frac{e^{-j\beta r_3}}{r_3} - f(\theta, \phi) \frac{e^{-j\beta r_4}}{r_4}$$

$$= [e^{j\beta s \cos \psi_1} - e^{j\beta s \cos \psi_2} + e^{j\beta s \cos \psi_3} - e^{j\beta s \cos \psi_4}] f(\theta, \phi) \frac{e^{-j\beta r}}{r} \quad (2)$$

where

$$\begin{aligned} \cos \psi_1 &= \hat{a}_x \cdot \hat{a}_r = \sin \theta \cos \phi \\ \cos \psi_2 &= \hat{a}_y \cdot \hat{a}_r = \sin \theta \sin \phi \\ \cos \psi_3 &= -\hat{a}_x \cdot \hat{a}_r = -\sin \theta \cos \phi \\ \cos \psi_4 &= -\hat{a}_y \cdot \hat{a}_r = -\sin \theta \sin \phi \end{aligned} \quad (3)$$

since

$$\hat{a}_r = \hat{a}_x \sin \theta \cos \phi + \hat{a}_y \sin \theta \sin \phi + \hat{a}_z \cos \theta \quad (3a)$$

Substituting (3) into (2) we can write it as

$$E^t(r, \theta, \phi) = 2 [\cos(\beta s \sin \theta \cos \phi) - \cos(\beta s \sin \theta \sin \phi)] f(\theta, \phi) \frac{e^{-j\beta r}}{r} \quad (4)$$

By letting the field of the isolated element (in this case the dipole) to be

$$E_\theta^o = f(\theta, \phi) \frac{e^{-j\beta r}}{r} = j\eta \frac{\beta I_0 l e^{-j\beta r}}{4\pi r} \sin \theta \quad (5)$$

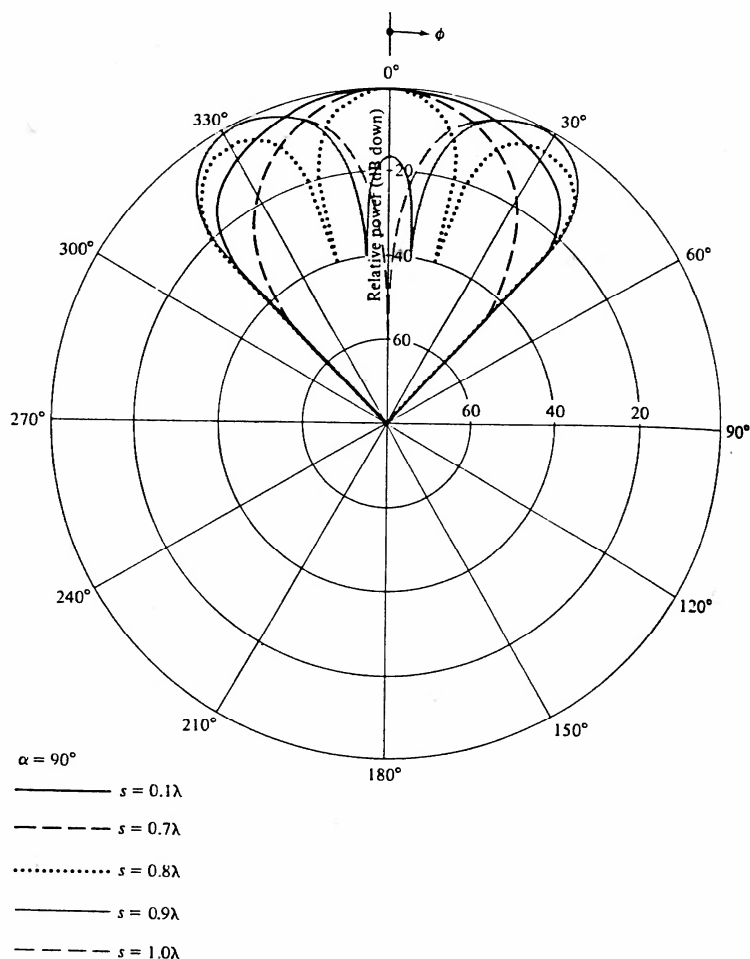
then (4) can be written as

$$E^t(r, \theta, \phi) = E_\theta^o F(\beta s)$$

where

$$F(\beta s) = 2 [\cos(\beta s \sin \theta \cos \phi) - \cos(\beta s \sin \theta \sin \phi)]$$

7.7



7.9 The period of each is

$$\alpha = 60^\circ: \Delta s = 2.0\lambda$$

$$\alpha = 45^\circ: \Delta s = 16.69\lambda$$

$$\alpha = 30^\circ: \Delta s = 30.0\lambda$$

while the peak values are

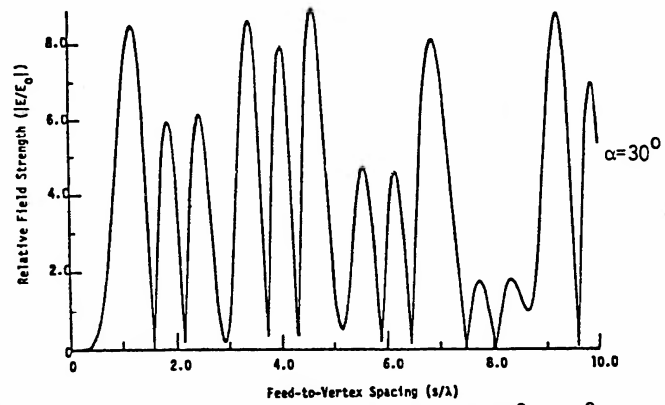
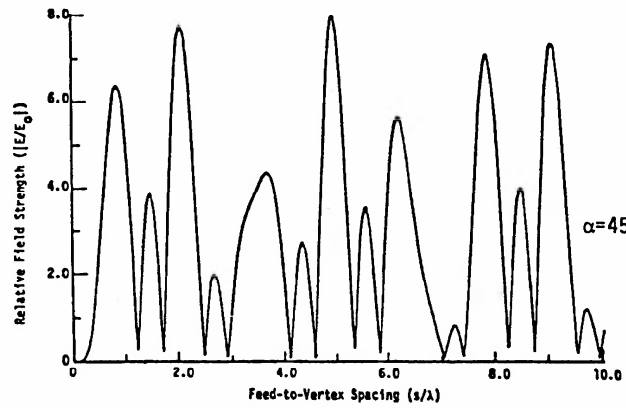
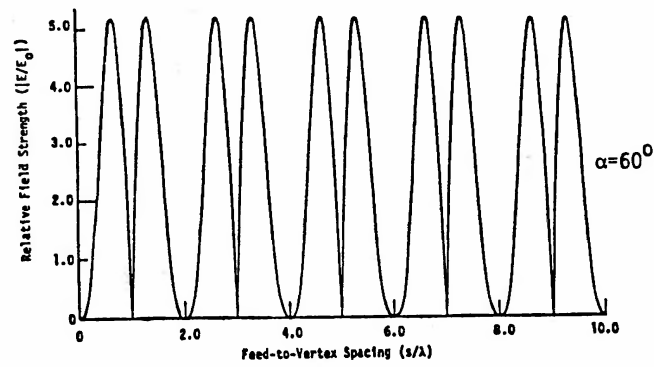
$$\alpha = 60^\circ: |E/E_0|_{\max} = 5.2$$

$$\alpha = 45^\circ: |E/E_0|_{\max} = 8.0$$

$$\alpha = 30^\circ: |E/E_0|_{\max} = 9.0$$

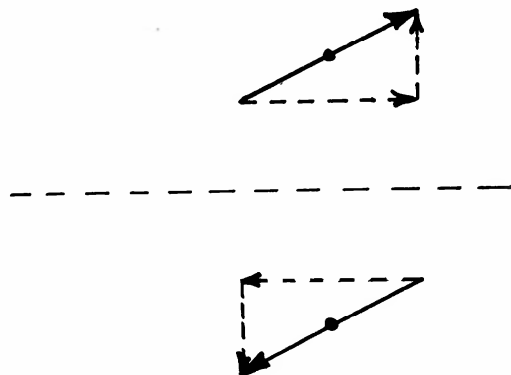
The plots of $|E/E_0|$ as a function of s in the range $0 \leq s \leq 10\lambda$

7.9 cont'd.



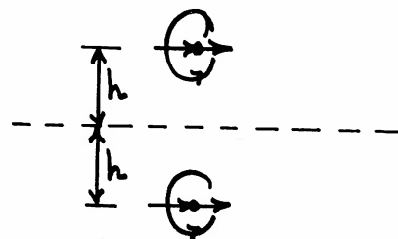
Relative field strengths along the axis ($\theta = 90^\circ$, $\phi = 0^\circ$) for $\alpha = 60^\circ$, 45° , 30° corner reflectors as a function of feed-to-vertex spacing.

7.10



7.11

Since a small loop can be modeled as a linear magnetic dipole, as shown in the figure, then using the sources and images of Figure 7-2(a) we choose the image to be another



magnetic horizontal linear dipole as shown in the figure. The corresponding small loop to represent the image is also shown.

7.13

Since $\underline{E}_a = \hat{a}_\rho E_\rho + \hat{a}_\phi E_\phi$ exists only over the circular aperture where $\rho' \leq a$, and it is zero elsewhere, then over the aperture and outside it along a flat plane coincident with the circular aperture we can form the electric and magnetic equivalent current, as shown below

$$\underline{M}_a = \hat{n} \times \underline{E}_a \quad \underline{M}_s = 0, \underline{J}_s \neq 0 \quad \underline{J}_s, \underline{M}_a \uparrow \hat{n} \quad \underline{M}_s = 0, \underline{J}_s \neq 0 \quad \epsilon_0, \mu_0$$

$\leftarrow a \quad a \rightarrow$

Now if we move a perfect electric conductor to coincide with the interface we have the following equivalent.

$$\underline{M}_s = \underline{J}_s = 0 \quad \underline{J}_s = 0, \underline{M}_a \uparrow \hat{n} \quad \underline{M}_s = \underline{J}_s = 0 \quad \epsilon_0, \mu_0$$

$\sigma = \infty$

which is equivalent to

$$\underline{M}_s = 2\hat{n} \times \underline{E}_a \quad \underline{J}_s = \underline{M}_s = 0 \quad \underline{J}_s = 0, \underline{M}_s = 2\underline{M}_a \quad \underline{J}_s = \underline{M}_s = 0 \quad \epsilon_0, \mu_0$$

$\leftarrow a \quad a \rightarrow$

$$\underline{M}_s = 2\hat{a}_\phi \times (\hat{a}_\rho E_\rho + \hat{a}_\phi E_\phi) = 2\hat{a}_\phi E_\rho - 2\hat{a}_\rho E_\phi$$

$$\boxed{7.14} \quad \underline{E}^i = \hat{a}_x E_0 e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{H}^i = \frac{E_0}{\eta} (-\hat{a}_y \cos \theta_i - \hat{a}_z \sin \theta_i) e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

a. Induction Equivalent

$$\underline{M}_i = 2 \hat{n} \times \underline{E}^i \Big|_{\substack{z=0 \\ y=y'}} = 2 \hat{a}_z \times \hat{a}_x E_x \Big|_{\substack{z=0 \\ y=y'}} = \hat{a}_y 2 E_0 e^{-j\beta_0 y' \sin \theta_i}$$

$$M_x = M_z = 0, \quad M_y = 2 E_0 e^{-j\beta_0 y' \sin \theta_i}$$

$$\text{Using (6-125a)-(6-125d)}$$

$$N_\theta = N_\phi = 0$$

$$L_\theta = \iint_{S_a} M_y \cos \theta_s \sin \phi_s e^{j\beta_0 y' \cos \psi} ds' = 2 E_0 \cos \theta_s \sin \phi_s \iint_{S_a} e^{-j\beta_0 y' \sin \theta_i} e^{j\beta_0 y' \cos \psi} ds'$$

Using (6-130b), (6-132a) and (6-132b) we can write

$$r' \cos \psi = x' \sin \theta_s \cos \phi_s + y' \sin \theta_s \sin \phi_s$$

$$ds' = dx' dy' = \rho' d\rho' d\phi'$$

Thus

$$L_\theta = 2 E_0 \cos \theta_s \sin \phi_s \iint_{S_a} e^{j\beta_0 [x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i)]} \rho' d\rho' d\phi'$$

$$\text{Since } \begin{aligned} x' &= \rho' \cos \phi' \\ y' &= \rho' \sin \phi' \end{aligned}$$

then the exponent of the exponential can be written as

$$x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i) = \rho' \left[\underbrace{\sin \theta_s \cos \phi_s \cos \phi'}_A + \sin \phi' \underbrace{(\sin \theta_s \sin \phi_s - \sin \theta_i)}_B \right]$$

$$= \rho' (A \cos \phi' + B \sin \phi')$$

$$= \rho' \sqrt{A^2 + B^2} \left(\frac{A}{\sqrt{A^2 + B^2}} \cos \phi' + \frac{B}{\sqrt{A^2 + B^2}} \sin \phi' \right)$$

$$= \rho' \sqrt{A^2 + B^2} (\cos \phi_0 \cos \phi' + \sin \phi_0 \sin \phi')$$

$$x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i) = \sqrt{A^2 + B^2} \rho' \cos(\phi' - \phi_0)$$

$$\text{where } A = \sin \theta_s \cos \phi_s$$

$$B = \sin \theta_s \sin \phi_s - \sin \theta_i$$

$$\phi_0 = \tan^{-1} \left(\frac{B}{A} \right)$$

7.14 Cont'd. Thus we can write L_θ as

$$\begin{aligned} L_\theta &= 2E_0 \cos\theta_s \sin\phi_s \int_0^a \left\{ \int_0^{2\pi} e^{j\beta_0 \sqrt{A^2+B^2} \rho' \cos(\phi' - \phi_0)} d\phi' \right\} \rho' d\rho' \\ &= 2E_0 \cos\theta_s \sin\phi_s \int_0^a \left[2\pi J_0(\beta_0 \sqrt{A^2+B^2} \rho') \right] \rho' d\rho' = 4\pi E_0 \cos\theta_s \sin\phi_s \int_0^a J_0(\beta_0 \sqrt{A^2+B^2} \rho') \rho' d\rho' \\ L_\theta &= 4\pi E_0 \cos\theta_s \sin\phi_s \left[a^2 \frac{J_1(\beta_0 \sqrt{A^2+B^2} a)}{\beta_0 \sqrt{A^2+B^2}} \right] = 4\pi a^2 E_0 \cos\theta_s \sin\phi_s \frac{J_1(\beta_0 \sqrt{A^2+B^2} a)}{\beta_0 \sqrt{A^2+B^2}} \end{aligned}$$

Using (6-125d)

$$L_\phi = \int_{S_a} M_y \cos\phi_s e^{j\beta_0 r' \cos\psi} ds' = 4\pi a^2 E_0 \cos\phi_s \frac{J_1(\beta_0 \sqrt{A^2+B^2} a)}{\beta_0 \sqrt{A^2+B^2}}$$

The electric and magnetic scattered fields can be written using (6-122a) - (6-122f) as

$$\begin{aligned} E_r^s &\approx H_\theta^s = 0 \\ E_\theta^s &\approx -j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\phi + \eta K_\theta^s) = -j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos\phi_s \frac{J_1(\beta_0 \sqrt{A^2+B^2} a)}{\beta_0 \sqrt{A^2+B^2}} \\ E_\phi^s &\approx j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\theta - \eta K_\phi^s) = j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos\theta_s \sin\phi_s \frac{J_1(\beta_0 \sqrt{A^2+B^2} a)}{\beta_0 \sqrt{A^2+B^2}} \\ H_\theta^s &\approx -\frac{E_\phi^s}{\eta}, \quad H_\phi^s = \frac{E_\theta^s}{\eta} \end{aligned}$$

For backscatter observations

$$\theta_s = \theta_i, \quad \phi_s = 3\pi/2$$

$$A = \sin\theta_s \cos\phi_s = 0$$

$$B = \sin\theta_s \sin\phi_s - \sin\theta_i = -\sin\theta_i - \sin\theta_i = -2\sin\theta_i \quad \left. \vphantom{B = \sin\theta_s \sin\phi_s - \sin\theta_i} \right\} \sqrt{A^2+B^2} = 2\sin\theta_i$$

Thus

$$E_\theta^s \approx 0$$

$$E_\phi^s \approx -j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos\theta_i \frac{J_1(2\beta_0 a \sin\theta_i)}{2\beta_0 a \sin\theta_i}$$

$$H_\theta^s \approx -\frac{E_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{E_\phi^s}{\eta}$$

7.14 Cont'd.

b. Physical Equivalent

$$\underline{J}_p = 2 \hat{n} \times \underline{H}^i \Big|_{z=0} = 2 \hat{a}_z \times (\hat{a}_y H_y^i + \hat{a}_z H_z^i) \Big|_{\substack{z=0 \\ y=y'}} = -\hat{a}_x 2 H_y^i \Big|_{\substack{z=0 \\ y=y'}} = \hat{a}_x 2 \frac{E_0}{\eta} \cos \theta_i e^{-j\beta_0 y' \sin \theta_i}$$

$$J_y = J_z = 0, \quad J_x = \frac{2E_0}{\eta} \cos \theta_i e^{-j\beta_0 y' \sin \theta_i}$$

Using (6-125a)-(6-125d) and the results of the induction equivalent, we can write

$$N_\theta = \iint_{S_a} J_x \cos \theta_s \cos \phi_s e^{j\beta_0 r' \cos \psi} ds' = \frac{4\pi a^2 E_0}{\eta} \cos \theta_i \cos \theta_s \cos \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$N_\phi = \iint_{S_a} -J_x \sin \phi_s e^{j\beta_0 r' \cos \psi} ds' = -\frac{4\pi a^2 E_0}{\eta} \cos \theta_i \sin \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$A = \sin \theta_s \cos \phi_s, \quad B = \sin \theta_s \sin \phi_s - \sin \theta_i$$

Now using (6-122a)-(6-122f) we can write the scattered electric and magnetic fields as

$$\underline{E}_r^s \approx H_\theta^s \approx 0$$

$$\underline{E}_\theta^s \approx -j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (\hat{r} \times \hat{\phi} + \eta N_\theta) = -j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos \theta_i \cos \theta_s \cos \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$\underline{E}_\phi^s \approx j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (\hat{r} \times \hat{\theta} - \eta N_\phi) = j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos \theta_i \sin \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$H_\theta^s \approx -\frac{\underline{E}_\phi^s}{\eta}, \quad H_\phi^s \approx \frac{\underline{E}_\theta^s}{\eta}$$

For backscatter observations

$$\theta_s = \theta_i, \quad \phi_s = \pi/2$$

$$\left. \begin{aligned} A &= 0 \\ B &= -2 \sin \theta_i \end{aligned} \right\} \sqrt{A^2 + B^2} = 2 \sin \theta_i$$

$$\underline{E}_\theta^s \approx 0$$

$$\underline{E}_\phi^s \approx -j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos \theta_i \frac{J_1(2\beta_0 a \sin \theta_i)}{2\beta_0 a \sin \theta_i}$$

$$H_\theta^s \approx -\frac{\underline{E}_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{\underline{E}_\phi^s}{\eta}$$

$$\boxed{7.15} \quad \underline{H}^i = \hat{a}_x H_0 e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{E}^i = \eta H_0 (\hat{a}_y \cos \theta_i + \hat{a}_z \sin \theta_i) e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

a. Induction Equivalent

$$\underline{M}_L = 2 \hat{n} \times \underline{E}^i \Big|_{y=y', z=0} = 2 \hat{a}_z \times (\hat{a}_y E_y + \hat{a}_z E_z) \Big|_{y=y', z=0} = -\hat{a}_x E_y \Big|_{y=y', z=0} = -\hat{a}_x 2\eta H_0 \cos \theta_i e^{-j\beta_0 y' \sin \theta_i}$$

$$M_y = M_z = 0, \quad M_x = -2\eta H_0 \cos \theta_i e^{-j\beta_0 y' \sin \theta_i}$$

Using (6-125a)-(6-125d) and the solution of Problem 7.14, we can write

$$L_\theta = \iint_{S_a} M_x \cos \theta_s \cos \phi_s e^{j\beta_0 r' \cos \psi} ds' = -4\pi a^2 \eta H_0 \cos \theta_i \cos \theta_s \cos \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$L_\phi = \iint_{S_a} -M_x \sin \phi_s e^{j\beta_0 r' \cos \psi} ds' = +4\pi a^2 \eta H_0 \cos \theta_i \sin \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$A = \sin \theta_s \cos \phi_s, \quad B = \sin \theta_s \sin \phi_s - \sin \theta_i$$

Now using (6-122a)-(6-122f) we can write the scattered electric and magnetic fields as

$$\underline{E}_r^s \approx H_r^s \approx 0$$

$$\underline{E}_\theta^s \approx -\frac{j\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\phi + \eta H_\phi^s) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos \theta_i \sin \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$\underline{E}_\phi^s \approx j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\theta - \eta H_\theta^s) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos \theta_i \cos \theta_s \cos \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$H_\theta^s \approx -\frac{E_\phi^s}{\eta}, \quad H_\phi^s \approx \frac{E_\theta^s}{\eta}$$

For backscatter observations

$$\theta_s = \theta_i, \quad \phi_s = 3\pi/2$$

$$\left. \begin{aligned} A &= 0 \\ B &= -2 \sin \theta_i \end{aligned} \right\} \sqrt{A^2 + B^2} = 2 \sin \theta_i$$

$$\underline{E}_\theta^s \approx +j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos \theta_i \frac{J_1(2\beta_0 a \sin \theta_i)}{2\beta_0 a \sin \theta_i}$$

$$\underline{E}_\phi^s \approx 0$$

$$H_\theta^s \approx -\frac{E_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{E_\theta^s}{\eta}$$

cont'd.

b. Physical Equivalent

$$\underline{J}_p = 2\hat{n} \times \underline{H}^i \Big|_{z=0} = 2\hat{a}_z \times \hat{a}_x H_x = \hat{a}_y 2H_0 e^{-j\beta_0 y' \sin\theta_i}$$

$$J_x = J_z = 0, \quad J_y = 2H_0 e^{-j\beta_0 y' \sin\theta_i}$$

Using (6-125a) - (6-125d) and the solution of Problem 7.14, we can write

$$\left. \begin{aligned} N_\theta &= \iint_{S_a} J_y \cos\theta_s \sin\phi_s e^{j\beta_0 r' \cos\psi} ds' = 4\pi a^2 H_0 \cos\theta_s \sin\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}} \\ N_\phi &= \iint_{S_a} J_y \cos\phi_s e^{j\beta_0 r' \cos\psi} ds' = 4\pi a^2 H_0 \cos\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}} \end{aligned} \right\}$$

$$A = \sin\theta_s \cos\phi_s, \quad B = \sin\theta_s \sin\phi_s - \sin\theta_i$$

Now using (6-122a) - (6-122f) we can write the scattered electric and magnetic fields as

$$\underline{E}_\theta^s \approx -j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (\underline{L}_\theta^s + \eta N_\theta) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos\theta_s \sin\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}}$$

$$\underline{E}_\phi^s \approx j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (\underline{L}_\phi^s - \eta N_\phi) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}}$$

$$H_\theta^s \approx -\frac{\underline{E}_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{\underline{E}_\theta^s}{\eta}$$

For backscatter directions

$$\theta_s = \theta_i, \quad \phi_s = \pi/2$$

$$\left. \begin{aligned} A &= 0 \\ B &= -2\sin\theta_i \end{aligned} \right\} \sqrt{A^2+B^2} = 2\sin\theta_i$$

$$\underline{E}_\theta^s \approx j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos\theta_i \frac{J_1(2\beta_0 a \sin\theta_i)}{2\beta_0 a \sin\theta_i}$$

$$\underline{E}_\phi^s \approx 0$$

$$H_\theta^s \approx -\frac{\underline{E}_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{\underline{E}_\theta^s}{\eta}$$