

Sample Midterm II with
Solutions

Name _____

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UNIVERSITY OF UTAH
ELECTRICAL & COMPUTER ENGINEERING DEPARTMENT

ANTENNA THEORY AND DESIGN

ECE 5324/6324

MIDTERM II

March 13, 2009

1. (25 points)

A half-wave dipole (antenna #1) is located at a distance of 0.25λ from a short circuited dipole (antenna #2) as sketched in Fig. 1:

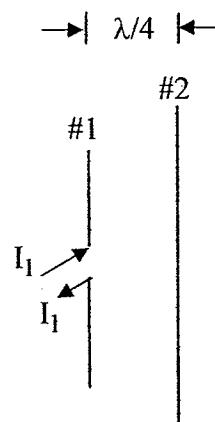


Fig. 1

It is given that $Z_{11} = 73 + j42$, $Z_{22} = 80 + j70$, and $Z_{12} = 38 - j30$ ohms.

- pts
- 10 a. Calculate the current induced in antenna #2 in terms of current fed into antenna #1 i.e. I_2 / I_1 (magnitude and phase).
- 10 b. Calculate the feed point impedance Z_1 of antenna #1 which is now different from Z_{11} .
- 5 c. What would be the feed point impedance Z'_1 of the driven antenna #1 if it were folded on itself to form a two-arm folded dipole of equal diameters for the two arms, respectively.

Midterm II - Solutions

4
March 13, 2009

1. a. $V_1 = I_1 Z_{11} + I_2 Z_{12}$ (1) $Z_{11} = 73 + j42$

$V_2 = 0 = I_1 Z_{12} + I_2 Z_{22}$ (2) $Z_{22} = 80 + j70$

$Z_{12} = 38 - j30$

From Eq. (2)

$$\begin{aligned}\frac{I_2}{I_1} &= -\frac{Z_{12}}{Z_{22}} = -\frac{(38 - j30)}{80 + j70} = -0.455 \angle \frac{-38.29^\circ}{41.18^\circ} \\ &= 0.455 \angle \frac{-79.47^\circ + 18^\circ}{^\circ} \\ &= 0.455 \angle \frac{+100.53^\circ}{^\circ} = -0.0832 \angle +j0.447\end{aligned}$$

b. From (1)

$$\begin{aligned}Z_1 &= \frac{V_1}{I_1} = Z_{11} + 0.455 e^{j100.53^\circ} (38 - j30) \\ &= 73 + j42 + 0.455 \times 48.41 e^{+j62.24^\circ} \\ &= 73 + j42 + 22.03 e^{j36.43^\circ} \\ &= 73 + j42 + 10.26 + j19.49 + j36.43^\circ \\ &= \boxed{83.26 + j61.49} = \boxed{103.50 e^{j36.43^\circ}}\end{aligned}$$

c. $Z'_1 = 4Z_1 = 414 e^{j36.43^\circ}$ or $4(83.26 + j61.49)$
 $= 333.04 + j245.96 \Omega$

2. (25 points)

pts

- 15 a. Design a **three** reactance network to match an antenna of impedance $Z_A = 60 + j30$ ohms to a two-wire transmission line of characteristic impedance $Z_0 = 350$ ohms using a circuit arrangement shown in Fig. 2.

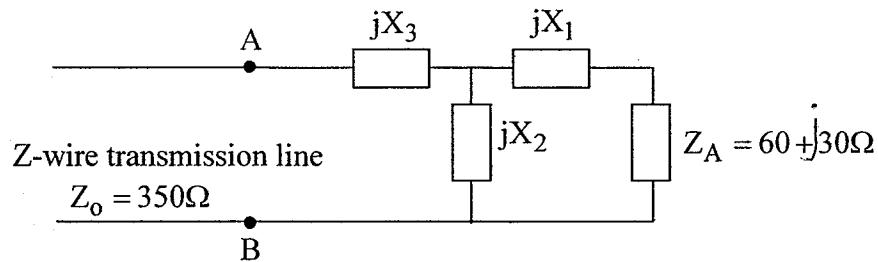


Fig. 2.

Select X_1 and X_2 such that

$$X_1 + X_2 = -X_A = -30\Omega$$

If possible, take X_1 to be a negative (capacitive) reactance.

- 10 b. For an antenna frequency of 400 MHz, calculate the values of three reactive elements, inductances and/or capacitances needed for your designed circuit.

2. Following a procedure similar to that of the course writeup "A reactive three-element circuit for Antenna Matching", we can write

$$Z_{AB} = \frac{(60 + j30 + jX_1)jX_2}{60 + j(30 + X_1 + X_2)} + jX_3 = 350 + j0 \quad (1)$$

It is given that $X_1 + X_2 = -30$. Thus Eq. 1 can be written as

$$Z_{AB} = \frac{(60 - jX_2)jX_2}{60} + jX_3 = 350 + j0 \quad (2)$$

Equating real and imaginary parts on the two sides of Eq.(2),

we can write

Equating imaginary parts

$$\frac{60}{60} X_2 + X_3 = 0$$

$$X_3 = -X_2$$

Equating real parts

$$\frac{X_2^2}{60} = 350 \Rightarrow X_2 = \pm 144.9$$

(1)

For X_2 , we can a capacitive reactance i.e. $X_2 = -144.9 = -\frac{1}{\omega C_2}$

$$C_2 = \frac{1}{\omega \times 144.9} = \frac{1}{2\pi \times 4 \times 10^8 \times 144.9} \\ = 2.746 \text{ pF}$$

Note that $X_1 + X_2 = -30$

$$X_1 = -30 - X_2 = -30 + 144.9 \\ = 114.9 = \omega L_1$$

$$L_1 = \frac{114.9}{2\pi \times 4 \times 10^8} = 45.7 \text{ nH}$$

$$X_3 = -X_2 = +144.9 = \omega L_3 \\ L_3 = 57.6 \text{ nH}$$

(2)

If possible, take x_1 to be a negative (capacitive) reactance (2')

Alternative Solution

From (1) on p. 2, we should take

$$x_2 = +144.9$$

$$\text{Then } jx_1 = j(-30 - x_2) = -j174.9 \Omega = -\frac{j}{\omega C_1}$$

$$C_1 = \frac{\cancel{10^{12}}}{2\pi \times 4 \times 10^8 \times 174.9} = 2.27 \text{ pF} \\ = 2270 \text{ fF}$$

$$jx_2 = j144.9 = j\omega L_2$$

$$L_2 = \frac{144.9 \times \cancel{10^10}}{2\pi \times 4 \times 10^8} = 57.65 \text{ nH}$$

$$jx_3 = -jx_2 = -j144.9 = \frac{-j}{\omega C_3}$$

$$C_3 = \frac{1}{2\pi \times 4 \times 10^8 \times 144.9} = 2.746 \text{ pF}$$

3. (25 points)

pts

- 5 a. Write expressions for the radiated electric field and power density of a ten-element collinear antenna array of $\lambda/2$ elements mounted one above the other so that the **axis of the array is along the z-axis**. Assume that each of the antennas are fed with equal magnitude and in-phase currents and that the center-to-center spacing between the antennas is 0.8λ .
- 5 b. Calculate the direction/s of maximum radiation for this antenna array.
- 5 c. Calculate the main beam width or beam width between first nulls for the principal lobe of radiation of the antenna array.
- 5 d. Neglecting mutual impedance effects, calculate the gain of the antenna array relative to an isotropic radiator.
- 5 e. For a total radiated power of 10 KW, calculate the electric field strength created at a distance of 12 miles (19.2 km).

3. This is a z-oriented antenna array of ten half wave dipoles (3)

a.

From p. 24 of the class Notes

$$N_z = 10, \alpha_z = 0; d_z = 0.8\lambda, \beta d_z = 1.6\pi$$

$$|AF| = \frac{\sin(N_z \Psi_{z/2})}{\sin(\Psi_{z/2})} = \frac{\sin\{5(1.6\pi \cos\theta)\}}{\sin\{\frac{1}{2}(1.6\pi \cos\theta)\}} \quad (1)$$

From

$$|E_0| = j \frac{60 I_m}{\pi} F(\theta) e^{-j \frac{\beta L}{2} \theta} \quad \frac{\beta L}{2} = \frac{2\pi}{\lambda} \times \frac{\lambda}{4} = 1\lambda$$

for $\lambda/2$ dipole

$$\text{where } F(\theta) = \frac{\cos(\frac{\beta L}{2} \cos\theta) - \cos(\frac{\beta L}{2})}{\sin\theta}$$

$$= \frac{\cos(\pi/2 \cos\theta)}{\sin\theta}$$

power density

$$S = \frac{|E_0|^2}{2\eta} |AF|^2$$

b. For direction of maximum radiation $\theta = 90^\circ$

Thus the maximum power density is in the xy plane (plane perpendicular to the orientation of the array).

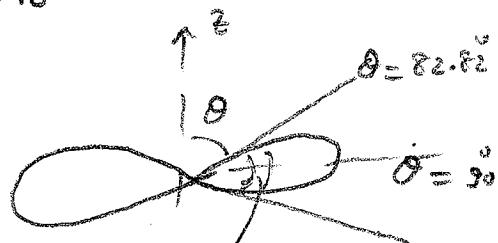
c. From Eq. (1) for directions of zero radiation

$$\frac{N_z \Psi_z}{2} = 8\pi \cos\theta_{FN} = \pm \pi$$

$$\theta_{FN} = \cos^{-1}\left(\frac{1}{8}\right) = 82.82^\circ$$

$$\rightarrow \cos^{-1}\left(-\frac{1}{8}\right) = 97.18^\circ$$

$$BWFN = 97.18^\circ - 82.82^\circ = 14.36^\circ$$



d. From Eq.(57) on p.33 of Notes

$$G = D = D_o N = 1.64 \times 10 = 16.4 \quad (12.15 \text{ dB}) \quad BWFN = 14.36^\circ$$

$$\frac{S_{max}}{S_0} = D S_0 = 16.4 \times \frac{P_{rad}}{4\pi r^2} = 16.4 \times \frac{10^4}{4\pi (19.2 \times 10^3)^2} = 3.54 \times 10^{-5} \text{ W/m}^2$$

3. (Continued)

(4)

$$\frac{E_{\max}}{\eta} = S = 3.54 \times 10^5$$

$$E_{\max} \left|_{rms} \right. = \sqrt{120 \pi \times 3.54 \times 10^5} = 115.52 \frac{mV}{m} \left|_{rms} \right.$$

$$E_{\max} \left|_{peak} \right. = 115.52 \sqrt{2} \frac{mV}{m} = 162.2 \frac{mV}{m} \left|_{peak} \right.$$

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Score:

Problem 1 _____ of a possible 25 points

Problem 2 _____ of a possible 25 points

Problem 3 _____ of a possible 25 points

Total _____ of a possible 75 points