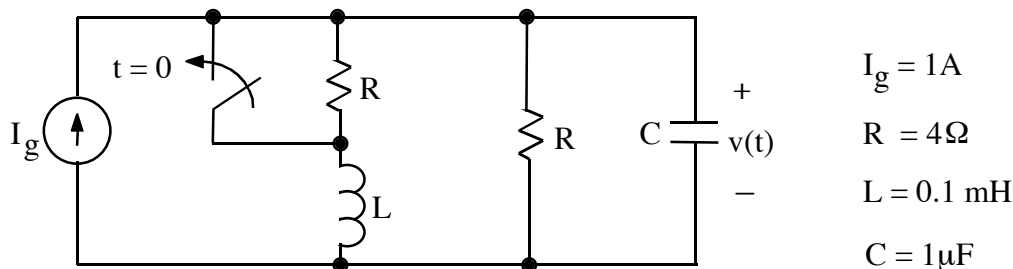


2. (45 points)



The current source is a dc current source. After being open for a long time, the switch is closed at $t = 0$.

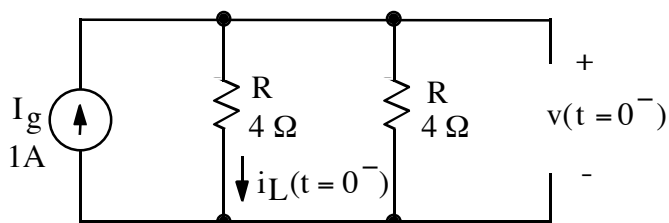
- Write a numerical time-domain expression for $v(t)$.
- From the Laplace transform of $v(t)$, find the numerical values of $v(t)$ for $t = 0^+$ and $t \rightarrow \infty$.

ans: a) $v(t > 0) = 2.003 e^{-0.4kt} - 0.003 e^{-249.6kt} \text{ V}$

b) $v(t = 0^+) = 2\text{ V}, v(t \rightarrow \infty) = 0\text{ V}$

sol'n: (a) First we find initial conditions for L and C. (We need these for s-domain models of L and C.)

For $t = 0^-$, L acts like short, C acts like open circuit.

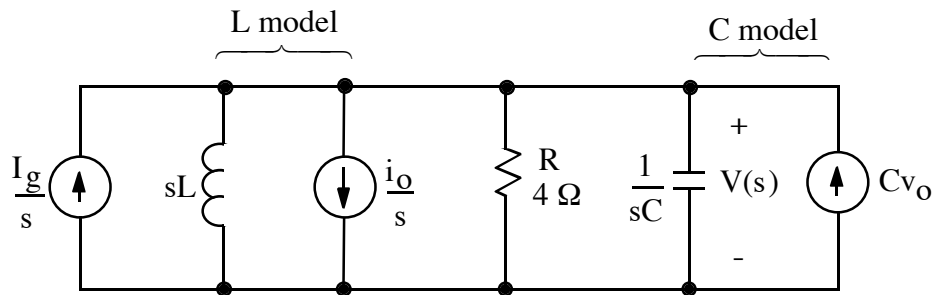


$$v(t = 0^-) = I_g \cdot R \parallel R = 1\text{ A} \cdot 2\Omega = 2\text{ V} \equiv v_o$$

$$i_L(t = 0^-) = I_g \cdot \frac{R}{R + R} = \frac{1}{2} \text{ A} \equiv i_o$$

When we close the switch, we short out the first R.

s-domain model:



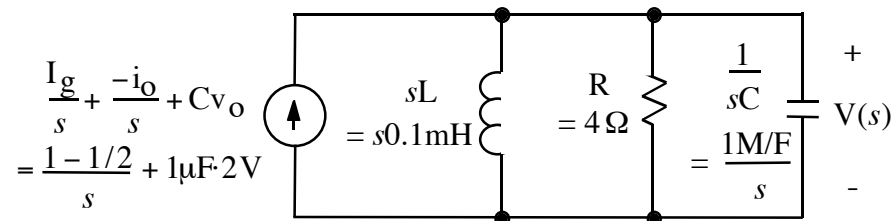
Note: A DC source corresponds to a step function even if there is no switch and the source has the same output for all time. Thus, we have I_g/s as the source in the s -domain. (Conceptually, we only need the current source for $t > 0$ because the initial conditions on L and C account for what the current source did for $t < 0$.)

$$\mathcal{L}\{I_g\} = \mathcal{L}\{I_g u(t)\} = \frac{1}{s}$$

Note: We may choose either a series sL and V -source for L or a parallel sL and I -source for L . Here, the parallel I -source model is more convenient. The same applies to the C .

Normally, we might use superposition at this point, turning on the I -sources one at a time and then summing currents or voltages to get a final answer.

Here, however, we have parallel I -sources that sum:



Combining the parallel impedances and using $V = I z$, we have

$$V(s) = \left(\frac{1}{2s} + 2\mu F V \right) \cdot sL \parallel R \parallel \frac{1}{sC}$$

To compute the parallel z value, we factor out $1/sC$ to remove fractions:

$$sL \parallel R \parallel \frac{1}{sC} = sL \parallel \frac{1}{sC} \parallel R = \frac{1}{sC} \cdot s^2 LC \parallel 1 \parallel sRC$$

Now multiply through by sC :

$$sL \parallel R \parallel \frac{1}{sC} = \frac{1}{sC} \cdot \frac{s^2 LC}{s^2 LC + 1} \parallel sRC = \frac{sL}{s^2 LC + 1} \parallel R$$

Factor out the denominator again:

$$sL \parallel R \parallel \frac{1}{sC} = \frac{1}{s^2 LC + 1} \cdot sL \parallel R(s^2 LC + 1)$$

$$sL \parallel R \parallel \frac{1}{sC} = \frac{1}{s^2 LC + 1} \cdot \frac{sLR(s^2 LC + 1)}{sL + R(s^2 LC + 1)}$$

Multiply through by $s^2 LC + 1$:

$$sL \parallel R \parallel \frac{1}{sC} = \frac{sLR}{sL + R(s^2 LC + 1)}$$

Now divide top and bottom by RLC to make the coefficient of the highest power of s in the denominator equal to unity.

$$sL \parallel R \parallel \frac{1}{sC} = \frac{s/C}{s^2 + \frac{1}{RC}s + \frac{1}{LC}}$$

Check: Using the numerator and the first term in the denominator, we have the following units analysis:

$$\frac{s/C}{s^2} = \frac{1}{sC}$$

Thus, we have an impedance as we should have. The other terms in the denominator have the same units as s^2 since the units of s are, ironically, 1/sec or 1/s.

Now we plug in numbers to compute $V(s)$:

$$V(s) = \left[\frac{1}{2s} + 2\mu \right] \parallel \left[\frac{\overset{\downarrow 1/C}{1M\cancel{\mu}}}{s^2 + \underbrace{\frac{1M}{4}s + 10G}_{\substack{\uparrow \frac{1}{RC} \quad \uparrow \frac{1}{LC}}}} \right]$$

quadratic poles term

Find poles for quadratic term in preparation for partial fractions:

$$s_{1,2} = -\frac{1M}{8} \pm \sqrt{\frac{1M^2}{8} - 10G}$$

||
 $\sqrt{(125k)^2 - (100k)^2}$

not complex poles

$$s_{1,2} = -125k \pm 75k \text{ rad/s} \quad (\text{based on } 5^2 - 4^2 = 3^2 \text{ pythagorean triple})$$

$$s_1 = -50k, s_2 = -200 \text{ rad/s}$$

Now use partial fractions:

$$V(s) = \frac{k_1}{s + 50k} + \frac{k_2}{s + 200k}$$

$$\begin{aligned} k_1 &= V(s)(s + 50k) \Big|_{s=-50k} = \frac{1 - 50k \cdot 4\mu}{2} \cdot \frac{1M}{-50k + 200k} \\ &= \frac{1 - 200m}{2} \cdot \frac{1M}{150k} \\ &= \frac{800\mu}{2} \cdot \frac{1M}{150k} = \frac{8}{3} \end{aligned}$$

$$\begin{aligned} k_2 &= V(s)(s + 50k) \Big|_{s=-200k} = \frac{1 - 200k4\mu}{2} \cdot \frac{1M}{-200k + 50k} \\ &= \frac{\cancel{m} \cdot 1M}{-2(150k)} = -\frac{2}{3} \end{aligned}$$

$$V(s) = \frac{8/3}{s + 50k} - \frac{2/3}{s + 200k}$$

Use the standard inverse Laplace transform term:

$$\mathcal{L}^{-1} \left\{ \frac{k}{s + a} \right\} = k e^{-at}$$

This gives the final answer:

$$v(t > 0) = 2.003 e^{-0.4kt} - 0.003 e^{-249.6kt} \text{ V}$$

sol'n: (b) Use the initial value theorem to find $v(t=0^+)$:

$$v(t = 0^+) = \lim_{s \rightarrow \infty} sV(s) = \lim_{s \rightarrow \infty} s \frac{1 + s4\mu}{2} \frac{1M}{s^2 + \frac{1M}{4}s + 10G}$$

The largest power of s dominates in the numerator and in the denominator.

$$v(t = 0^+) = \lim_{s \rightarrow \infty} sV(s) = \lim_{s \rightarrow \infty} \frac{s^2 4\mu 1M}{2s^2} = \frac{4}{2} = 2V \quad \checkmark$$

Note: We expect $v(0^+) = 2V$ since this is the initial capacitor voltage.

Use the final value theorem to find $v(t \rightarrow \infty)$:

$$\begin{aligned} v(t \rightarrow \infty) &= \lim_{s \rightarrow 0} sV(s) = \lim_{s \rightarrow 0} s \frac{1 + s4u}{2} \frac{1M}{s^2 + \frac{1M}{4}s + 10G} \\ &= 0 \cdot \frac{1}{2} \cdot \frac{1M}{10G} = 0V \quad \checkmark \end{aligned}$$

Note: We expect $v(t)$ to decay, since L becomes a short.