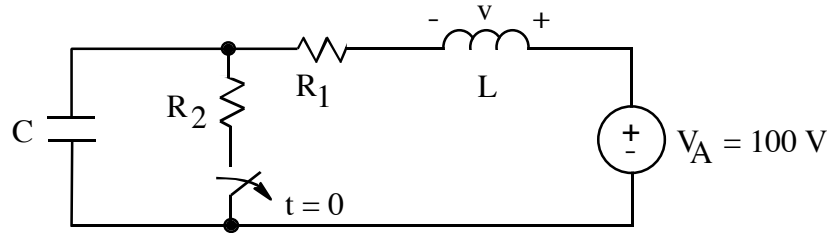


1. (50 points)



After having been open for a long time, the switch is closed at  $t = 0$ .

$$R_1 = 12.5\Omega \quad R_2 = 12.5\Omega \quad L = 6.25 \mu\text{H}$$

- Two capacitances are available: 250 nF and 2 nF. Specify the value of  $C$  that will make  $v(t)$  overdamped.
- Using the value of  $C$  found in (a), write a time-domain expression for  $v(t)$ .

**ans: a)**  $C = 250 \text{ nF}$

**b)**  $v(t) = 13.3 (e^{-0.4Mt} - e^{-1.6Mt}) \text{ V}$

**sol'n: (a)** To make the response overdamped, we must have two real characteristic roots. We use the circuit for  $t > 0$ , consisting of  $C$ ,  $R_1$ ,  $L$ , and  $v_A$  in series. We may find the characteristic equation by looking it up in a textbook or by setting the impedance of  $R_1$ ,  $C$ , and  $L$  in series to zero.

$$z = R_1 + \frac{1}{sC} + sL = s^2 + \frac{R_1}{L}s + \frac{1}{LC} = 0$$

The characteristic roots for the quadratic equation are

$$s_{1,2} = -\frac{R}{2L} \pm \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

or

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_o^2} \quad \alpha \equiv \frac{R}{2L} \quad \omega_o \equiv \frac{1}{\sqrt{LC}}$$

We want an overdamped response, (real roots  $\alpha^2 > \omega_o^2$ ).

$$\alpha = \frac{R}{2L} = \frac{12.5\Omega}{(2) 6.25 \mu\text{H}} = \frac{12.5}{12.5} \text{ M rad/s} = 1 \text{ M rad/s}$$

Try each  $C$  value in turn.

$C = 2 \text{ nF}$ :

$$\omega_o = \frac{1}{\sqrt{6.25 \mu\text{H} \cdot 2 \text{ nF}}} = \frac{1}{\sqrt{12.5 \text{ m} \cdot 1\mu}} = \frac{1\text{M}}{1.1118} \text{ rad/s}$$

$$\omega_o = 8.9\text{M rad/s} > \alpha^2 \text{ (underdamped)}$$

$C = 250 \text{ nF}$ :

$$\omega_o = \frac{1}{\sqrt{6.25 \mu\text{H} \cdot 250 \text{ nF}}} = \frac{1}{\sqrt{1562.5 \text{ m} \cdot 1\mu}} = \frac{1\text{M}}{1.25} \text{ rad/s}$$

$$\omega_o = 0.8\text{M rad/s} < \alpha^2 \text{ (overdamped)}$$

We need  $C = 250 \text{ nF}$  for an overdamped solution.

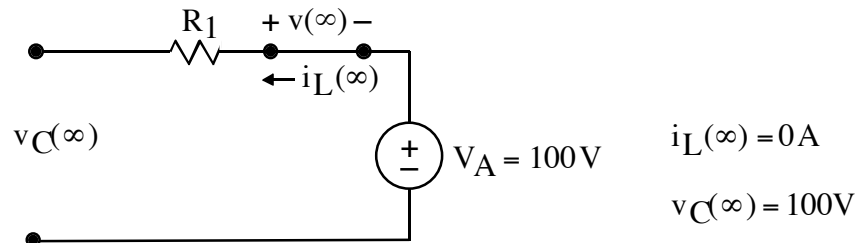
**sol'n: (b)** We use the exponential solution for the overdamped case:

$$v(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t} + A_3$$

Because the value of  $A_3$  is all that is left of  $v(t)$  as  $t \rightarrow \infty$ , we first find the constant term,  $A_3$ . (The other terms decay because the characteristic roots always have negative real parts in a passive RLC circuit. When the switch opens, the energy sloshing back and forth in the L and C will decay owing to power dissipated by the series resistor  $R_1$ .)

As  $t \rightarrow \infty$ , the circuit reaches equilibrium. C acts like an open circuit, L acts like a short circuit or wire.

Model:



Since L acts like a wire, there is no voltage drop across it.

Thus,  $A_3 = v(t \rightarrow \infty) = 0$ .

We find coefficients  $A_1$  and  $A_2$  by matching initial conditions in the circuit. We find initial conditions by examining the circuit at  $t = 0^-$ , when the circuit has reached equilibrium. We find the values of  $i_L$  and  $v_C$ , the energy variables, at  $t = 0^-$  and use the same values at  $t = 0^+$  (since the energy in the circuit cannot change instantly).

Mathematically, our general form of solution for the overdamped case gives the following values for  $v(0^+)$  and  $dv(t)/dt|_{t=0^+}$ :

$$v(0^+) = A_1 + A_2 + A_3 = A_1 + A_2$$

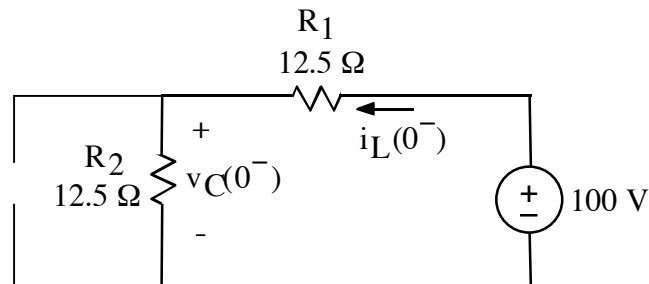
$$\left. \frac{dv(t)}{dt} \right|_{t=0^+} = A_1 s_1 + A_2 s_2.$$

Note: We must always differentiate first and then plug in  $t = 0^+$ . Otherwise, we always get zero.

Now we find the numerical values of  $v(0^+)$  and  $dv(t)/dt|_{t=0^+}$ .

At  $t = 0^-$ , C acts like an open circuit and L acts like a short circuit.

Model:



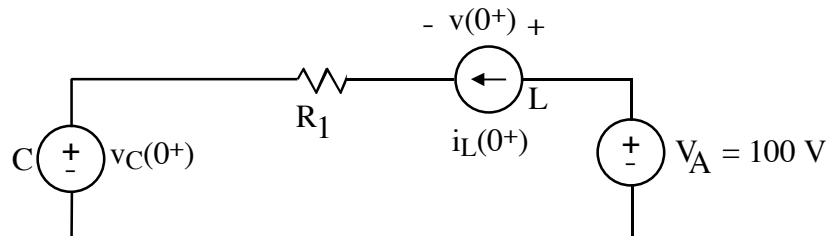
$$i_L(0^-) = \frac{100\text{V}}{25\Omega} = 4\text{A}$$

$$v_C(0^-) = 100\text{V} \cdot \frac{12.5\Omega}{25\Omega} = 50\text{V}$$

At time  $t = 0^+$ , we have  $i_L(0^+) = i_L(0^-) = 4\text{A}$  and  $v_C(0^+) = v_C(0^-) = 50\text{V}$ . We solve the circuit at  $t = 0^+$ , treating  $i_L(0^+)$  as a current source and  $v_C(0^+)$  as a voltage source.

We now solve for  $v(0^+)$  and  $dv(t)/dt|_{t=0^+}$ . From these we find  $A_1$  and  $A_2$ .

Model:



We may apply any standard method to solve the circuit, but we can solve the above circuit using a voltage loop.

$$v(0^+) = v_A - i_L(0^+)R_1 - v_C(0^+) = 100\text{V} - 4\text{A} \cdot 12.5\Omega - 50\text{V} = 0\text{ V}$$

The same equation applies for  $t > 0$ , and we may differentiate to find  $dv(t)/dt$  in terms of energy (or state) variables  $i_L$  and  $v_C$ .

$$v(t) = v_A - i_L(t)R_1 - v_C(t)$$

$$\frac{dv(t)}{dt} = -\frac{di_L(t)}{dt}R_1 - \frac{dv_C(t)}{dt}$$

The basic equations for L and C, rearranged, allow us to translate the derivatives on the right side of this equation into non-derivatives we can calculate numerically.

$$\frac{di_L(t)}{dt} = \frac{1}{L}v_L(t)$$

$$\frac{dv_C(t)}{dt} = \frac{1}{C}i_C(t)$$

Applying these identities, we have

$$\frac{dv(t)}{dt} = -\frac{1}{L}v_L(t)R_1 - \frac{1}{C}i_C(t).$$

Only now that we have differentiated do we finally evaluate the derivative we seek at  $t = 0$ :

$$\left. \frac{dv(t)}{dt} \right|_{t=0^+} = -\frac{1}{L}v_L(0^+)R_1 - \frac{1}{C}i_C(0^+).$$

$$\left. \frac{dv(t)}{dt} \right|_{t=0^+} = -\frac{1}{6.25\mu\text{H}} \cdot 0\text{V} \cdot 12.5\Omega - \frac{1}{250\text{nF}}i_C(0^+).$$

Since  $i_C$  is in series with  $i_L$ , we have  $i_C(0^+) = i_L(0^+) = 4\text{A}$ .

$$\left. \frac{dv(t)}{dt} \right|_{t=0^+} = -\frac{4\text{A}}{250\text{nF}} = -16\text{ MV/s}$$

Now we find  $A_1$  and  $A_2$ .

$$v(0^+) = 0 = A_1 + A_2 \Rightarrow A_2 = -A_1$$

$$\left. \frac{dv(t)}{dt} \right|_{t=0^+} = -16\text{MV/s} = A_1s_1 + A_2s_2 = A_1(s_1 - s_2).$$

$$\begin{aligned}
s_1 - s_2 &= -\alpha + \sqrt{\alpha^2 - \omega_0^2} - \left(-\alpha - \sqrt{\alpha^2 - \omega_0^2}\right) \\
&= 2\sqrt{\alpha^2 - \omega_0^2} \\
&= 2\sqrt{(1 M)^2 - (0.8 M)^2} \\
&= (2) 0.6 M = 1.2 M
\end{aligned}$$

Concluding the algebra, we find the numerical values of the coefficients  $A_1$  and  $A_2$ .

$$A_1 = \frac{16 M \text{ v/s}}{1.2 M} = 13.3 \text{ v/s}$$

$$A_2 = -13.3 \text{ v/s}$$

Using the values of  $\alpha$  and  $\omega_0$  from above, we find the values of  $s_1$  and  $s_2$ .

$$s_1 = -1 M + 0.6 M = -0.4 M$$

$$s_2 = -1 M - 0.6 M = -1.6 M$$

Plugging into the general form of underdamped solution completes our answer:

$$v(t) = 13.3 (e^{-0.4Mt} - e^{-1.6Mt}) \text{ V}$$