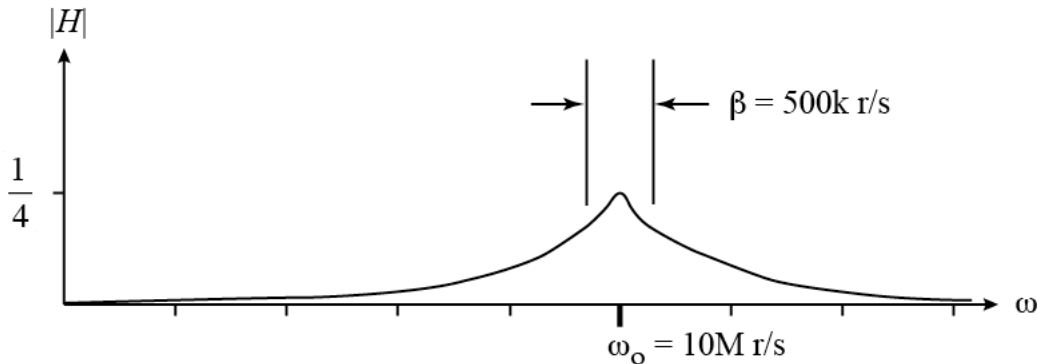
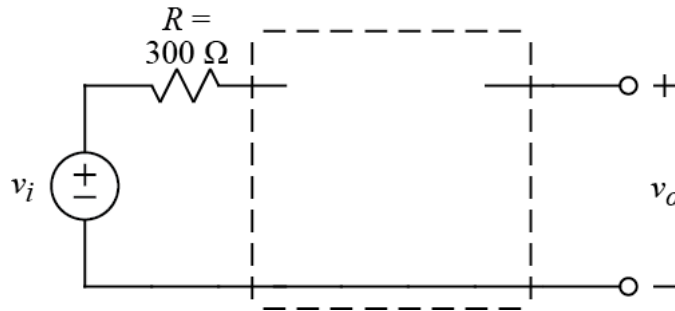


Ex:



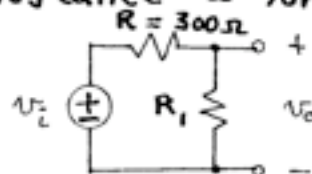
Given the resistor connected as shown and using not more than one each R , L , and C in the dashed-line box, design a circuit to go in the dashed-line box that will produce the **band-pass** $|H(j\omega)|$ vs. ω shown above. That is:

$$\max_{\omega} |H(j\omega)| = \frac{1}{4} \text{ and occurs at } \omega_0 = 10 \text{ M r/s}$$

The bandwidth, β , of the filter is 500k r/s.

$$|H(j\omega)| = 0 \text{ at } \omega = 0 \quad \text{and} \quad \lim_{\omega \rightarrow \infty} |H(j\omega)| = 0$$

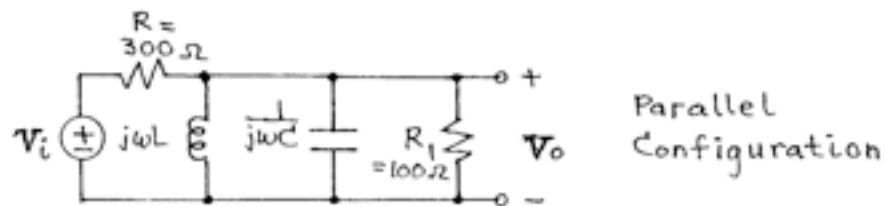
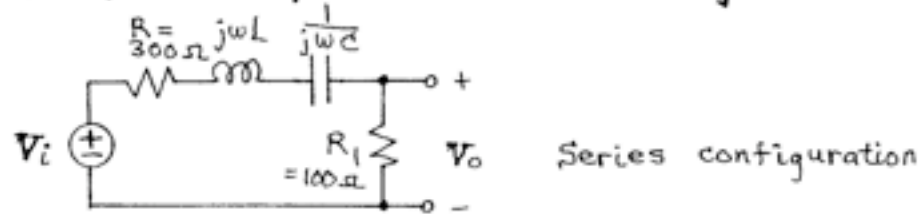
SOL'N: To achieve the peak at ω_0 , we may use a series LC in the top rail or a parallel LC from the top to bottom rail. To achieve a gain of $1/4$ at ω_0 , we must use a vertical resistance to form a V-divider with R .



$$\frac{R_1}{R + R_1} = \frac{1}{4}, \quad 3R_1 = R = 300 \Omega$$

$$R_1 = 100 \Omega$$

We have two possible circuit configurations:



Note: R_1 must be to the right of L and C in order to have any effect from the L and C in the series configuration.

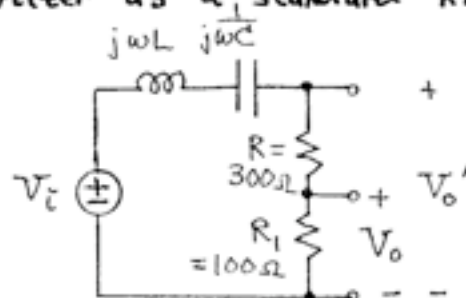
For the series configuration, the L and C are to act like a wire at $\omega_0 = 10 \text{ M r/s}$.

$$j\omega_0 L + \frac{1}{j\omega_0 C} = 0 \quad \text{or} \quad \omega_0^2 = (10 \text{ M r/s})^2 = \frac{1}{LC}$$

The bandwidth when using a series L and C is

$$\beta = \frac{R_{eq}}{L} = 500 \text{ kr/s}$$

To determine the value of R_{eq} , we view the filter as a standard RLC filter and a V-divider:



$$H(j\omega) = \frac{V_o'}{V_i} = \frac{V_o}{V_o'}$$

$$H(j\omega) = H'(j\omega) \cdot \frac{R_1}{R + R_1}$$

The cutoff frequencies for $H(j\omega)$ are the same as the cutoff frequencies for $H'(j\omega)$:

$$\omega_{c1,2} = \pm \frac{R_{eq}}{2L} \pm \sqrt{\left(\frac{R_{eq}}{2L}\right)^2 + \omega_0^2}, \quad \beta = \frac{R_{eq}}{L}$$

$$\text{where } R_{eq} = R + R_1 = 400 \Omega$$

Using $\beta = R_{eq}/L$, we find L :

$$L = \frac{R_{eq}}{\beta} = \frac{400 \Omega}{500 \text{kr/s}} = 0.8 \text{ mH or } 800 \mu\text{H}$$

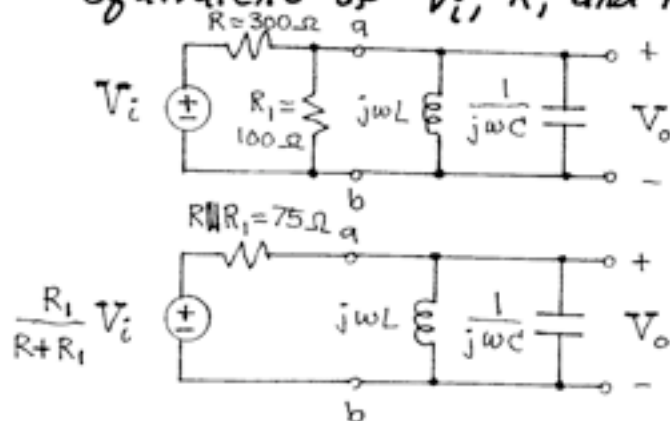
Using $\omega_0^2 = \frac{1}{LC}$ and $L = 800 \mu\text{H}$, we find C :

$$C = \frac{1}{\omega_0^2 L} = \frac{1 \text{ F}}{10\text{M} \cdot 10\text{M} \cdot 800 \mu} = \frac{1 \mu\text{F}}{80\text{K}}$$

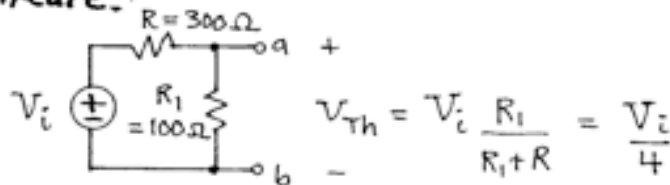
$$C = 12.5 \text{ pF}$$

Summary of series RLC: $R_1 = 100 \Omega$, $L = 800 \mu\text{H}$, $C = 12.5 \text{ F}$

For the parallel configuration, we move R_1 to the left of L and C and use a Thevenin equivalent of V_i , R_1 and R_1 .



To find the Thevenin equivalent, we find V_{Th} by finding the open-circuit ^{output} v of the V_i , R , and R_1 circuit.



To find R_{Th} , we turn off V_i and look in from terminals a and b. The resistance seen is

$$R_{Th} = R \parallel R_1 = 300 \Omega \parallel 100 \Omega = 75 \Omega$$

Using the filter with the Thevenin equivalent, we have

$$H(j\omega) = \frac{V_o}{V_i} = \frac{R_1}{R + R_1} \frac{V_o}{V_i'} = \frac{1}{4} H'(j\omega)$$

$$\text{where } H'(j\omega) \equiv \frac{V_o}{V_i'} \quad \text{where } V_i' \equiv \frac{R_1}{R + R_1} V_i$$

The cutoff frequencies of $H'(j\omega)$ are the same as the cutoff frequencies of $H(j\omega)$.

$$\omega_{c1,2} = \pm \frac{1}{2RC_{Th}} + \sqrt{\left(\frac{1}{2RC_{Th}}\right)^2 + \omega_0^2}, \quad \beta = \frac{1}{RC_{Th}}$$

$$\omega_0^2 = \frac{1}{LC}$$

Using R_{Th} and β , we find C :

$$C = \frac{1}{R\beta_{Th}} = \frac{1}{75 \Omega \cdot 500 \text{kr/s}} = 26.6 \text{ nF}$$

Using L and ω_0^2 , we find L :

$$L = \frac{1}{\omega_0^2 C} = \frac{1}{10 \text{M} / 10 \text{M} \cdot 26.6 \text{n}} = 375 \text{ nH}$$