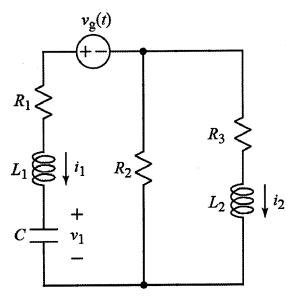
U

Ex:



At t = 0, $v_g(t)$ switches instantly from $-v_0$ to v_0 .

a) Write the state-variable equations for the circuit in terms of the state vector:

$$\vec{x} = \begin{bmatrix} i_1 \\ i_2 \\ v_1 \end{bmatrix}$$

- b) Evaluate the state vector at $t = 0^+$.
- soln: a) On the left of each egin, we have the first derivatives of energy vars. We equate these derivatives with non-derivatives via the component egins for L and C.

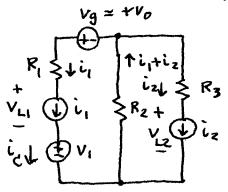
$$di_{1}/dt = V_{L1}/L_{1}$$

$$di_{2}/dt = V_{L2}/L_{2}$$

$$dv_{1}/dt = ic/C$$

Now we find the egins for expressing v_{L1} , v_{L2} , and i_{c} in terms of energy vars i_{1} , i_{2} , and v_{1} .

We may replace the L's and C with sources labeled i,, iz, and vc.



From Kirchhoff's laws, we have current $i_1 + i_2$ flowing up in the center branch, since i_1 and i_2 are flowing down in the outer branches.

For v_{LI} , we may use a v-loop on the left: (proceed clockwise from lower left)

$$v_1 + v_{L_1} + i_1 R_1 - v_0 - (i_1 + i_2) R_2 = 0V$$

or

 $v_{L_1} = v_0 + (i_1 + i_2) R_2 - v_1 - i_1 R_1$

For VLZ, we may use a V-loop on the right:

$$-(i_1+i_2)R_2-i_2R_3-V_{12}=0V$$

$$V_{12}=-(i_1+i_2)R_2-i_2R_3$$

Finally, we have
$$i_{c} = i_{1}$$
.

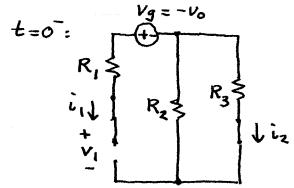
$$\frac{di_{1}}{dt} = \frac{v_{0} + (i_{1} + i_{2})R_{2} - v_{1} - i_{1}R_{1}}{L_{1}}$$

$$\frac{di_{2}}{dt} = \frac{-(i_{1} + i_{2})R_{2} - i_{2}R_{3}}{L_{2}}$$

$$\frac{dv_{1}}{dt} = \frac{i_{1}}{c}$$

b) The state vector, being energy vars, has the same value at t=0+ as at t=0-.

At t=0, we may model the L as a wire and the C as an open.



 $i_1(0^-) = 0 A$ since the C in series with the L is an open.

 $i_2(0^-) = 0A$ since there is no pwr source connected to $R_2//R_3$.

 $V_1(0^-) = -V_0$ since no current flows in R_1 or $R_2 \parallel R_3$ and there is no V-drop across the $R^1 s$.

$$\begin{bmatrix} i_1(o^+) \\ i_2(o^+) \\ v_1(o^+) \end{bmatrix} = \begin{bmatrix} o & A \\ o & A \\ -v_o \end{bmatrix}$$