

Microwave Engineering I
Final
April 29, 2002

Name Key

SS# _____

You may use your portfolio and a calculator.

Grading:

You will receive the highest of....

Midterm I	OR	Final Problems 1 and 2
Midterm II	OR	Final Problems 3 and 4
Midterm III	OR	Final Problems 5 and 6

Good luck and do well!

Problem 1 (50 points) Smith Charts, Lossy Lines, Double Stub Matching

(a) An antenna has an impedance of $25 + j10 \Omega$. It is connected to a 50Ω line that is 1 meter = 2.125λ long. The line has a loss factor of $\alpha = 2 \text{ Np/m}$. What is the input impedance to this line?

(b) Design a double stub matching network for a load that is ~~$10 - j10 \Omega$~~ ^{$35 + j5 \Omega$} . Stubs are parallel, short circuited lines. The distance between the two stubs is 0.125λ . All lines are 50Ω .

$$(a) Z_L = 25 + j10 \Omega / 50 \Omega = 0.5 + j0.2$$

$$e^{-2\alpha z} = 0.2$$

\uparrow \uparrow
 z l_m

$$Z_{in} = (1.05 + j0.15) 50 \Omega = 52.5 + j7.5 \Omega$$

$$(b) Z_L = \frac{35 - j5 \Omega}{50 \Omega} = 0.7 - j0.1 \quad \text{Draw circle } 0.125 \text{ TWG}$$

Convert to $y_L = 1.4 - j0.2$

Follow real circle to $y_A = 1.4 + j0.1$

Rotate TWG to matching circle $y_B = 1.0 - j0.325$

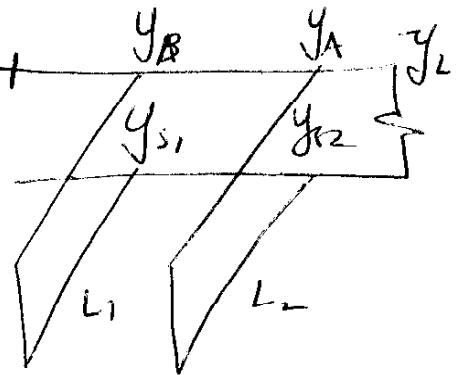
$$1 + j0 = y_{s1} + y_B \Rightarrow y_B = j0.325$$

$$y_A = y_{s2} + y_L \Rightarrow y_{s2} = j0.3$$

Rotate y_{s1} TWL to y_{sc}

$$L_1 = 0.25 - 0.05 = 0.2 \lambda$$

$$L_2 = 0.25 - 0.048 = 0.202 \lambda$$



Problem 2 (50 points) Steady State Transmission Lines

A 50Ω transmission line is fed by a 1V sinusoidal generator that has an impedance of 25Ω . The line is 0.3λ long. The load is $200 - j10 \Omega$.

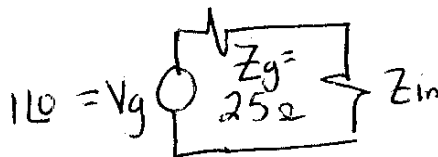
- Find the voltage standing wave ratio on the line.
- Write an expression for the positive traveling CURRENT on the line.
- Write a TIME DOMAIN expression for the TOTAL VOLTAGE on the line.

Note: If you do not have time to compute the complex values involved in this problem, write expressions for the items that you would use, including the values of all variables.

$$(a) Z_L = \frac{200 - j10}{50} = 4 - j0.2$$

$$\boxed{VSWR = 4.0}$$

$$(b) Z_{in} = (0.26 + j0.3) 50 \Omega =$$



$$V_{in} = \frac{Z_{in}}{Z_{in} + Z_g} V_g =$$

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0} =$$

$$V_{in} = V_0^+ e^{-j\beta L} + \Gamma_L V_0^+ e^{+j\beta L}$$

$$V_0^+ = \frac{V_{in}}{e^{-j\frac{2\pi}{\lambda} 0.3\lambda} + \Gamma_L e^{+j2\pi(0.3)}}$$

$$\boxed{I_0^+ = \frac{V_0^+}{Z_0} =}$$

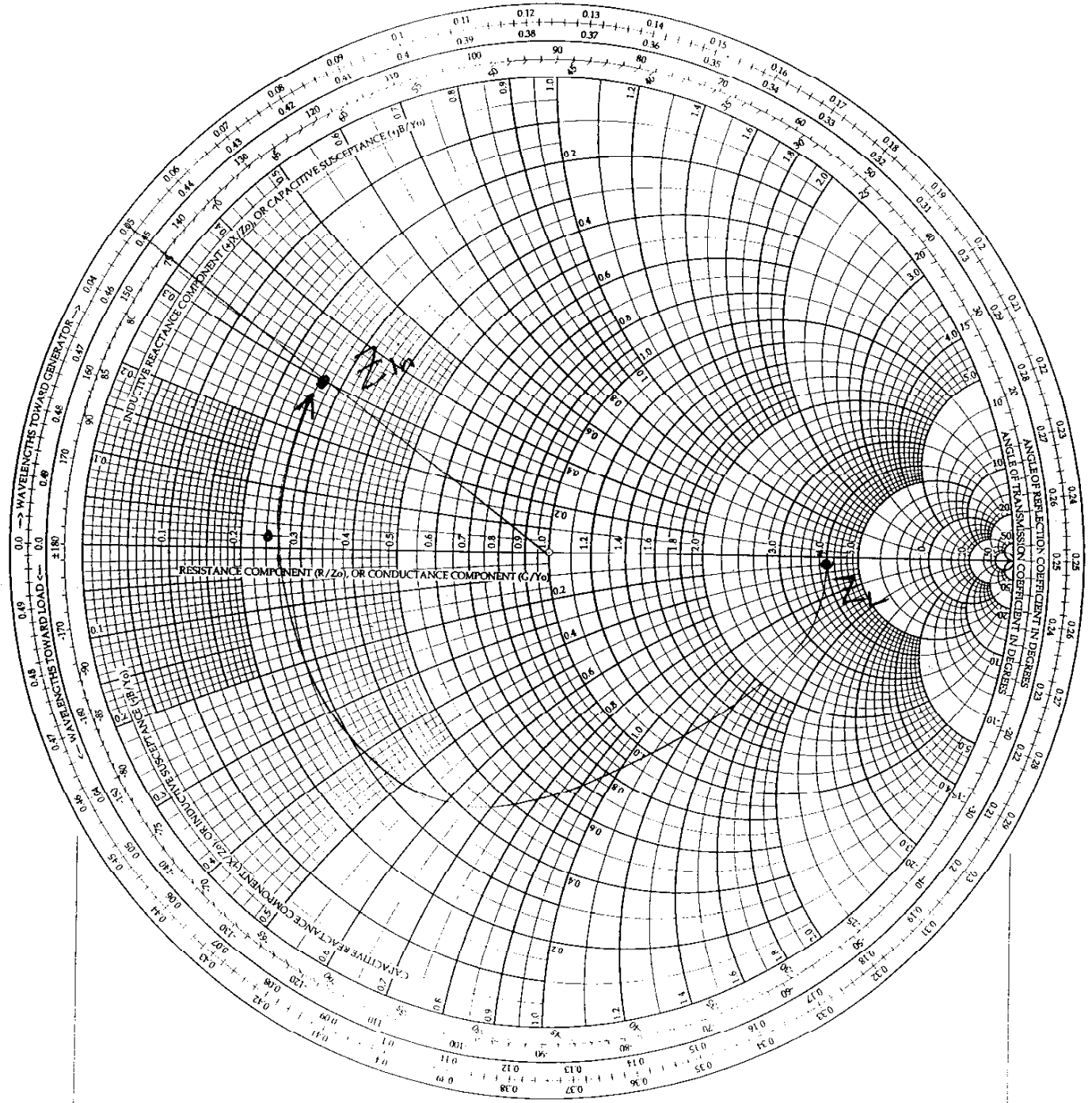
$$\tilde{V}(z) = V_0^+ e^{-j\beta z} + \Gamma_L V_0^+ e^{+j\beta z}$$

$$V(z, t) = |V_0^+| \cos(\omega t - \beta z + \phi_{V_0^+})$$

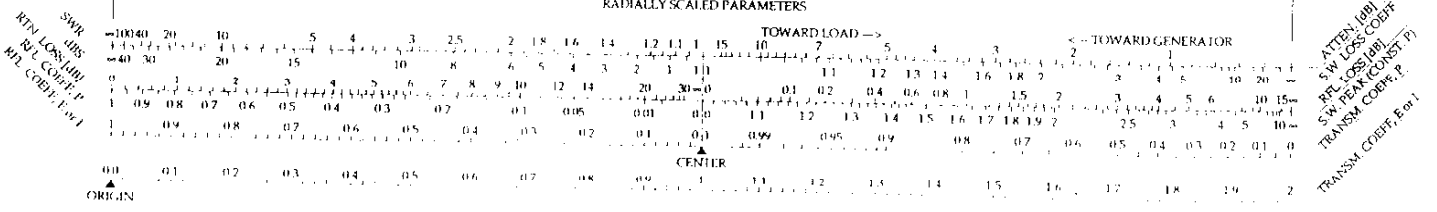
$$+ |\Gamma_L| |V_0^+| \cos(\omega t + \beta z + \phi_{V_0^+} + \phi_{\Gamma_L})$$

The Complete Smith Chart

Black Magic Design



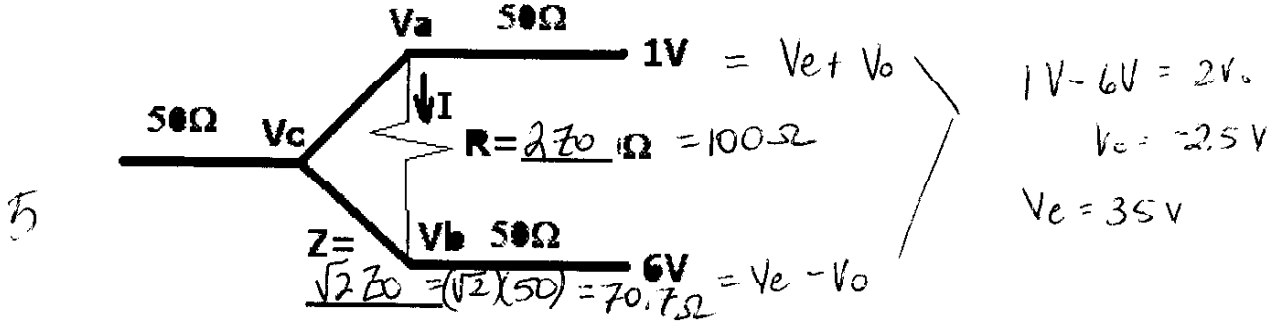
RADIALLY SCALED PARAMETERS



ATTEN (dB)
SW LOSS COEFF
dBS LOSS (dB)
SW LOSS (dB)
TRANSM COEFF, P
TRANSM COEFF, P_{avg}

Problem 3 (50 points) Couplers and Power Dividers

(a) Design a Wilkinson power divider for equal power combining. All input and output lines are 50Ω . Specify R and Z below.



(b) If the input voltages are 1 V and 6 V, find:

10 $V_a = \underline{\quad\quad} V = 1.75 + -1.25 = 0.5$

10 $V_b = \underline{\quad\quad} V = 1.75 + 1.25 = 3$

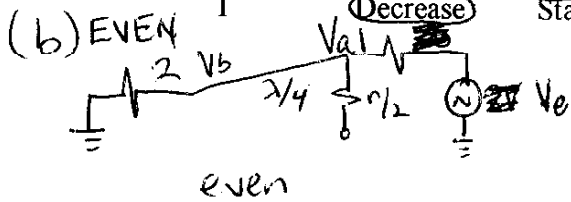
10 $V_c = \underline{\quad\quad} V = -j \frac{3.5}{\sqrt{2}} + 0 =$

10 $I = \underline{-25} \text{ mAmps} = 0 - .025 A$

(c) If a mistake is made in manufacturing, and a resistor that is twice as large as it should be ($=2R$) is installed, what will happen to each value:

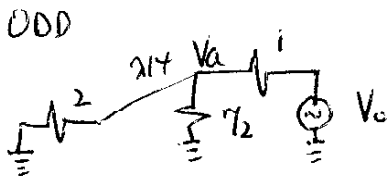
10

V_a	Decrease	Stay the same	Increase	V_a even stays same, V_a odd \uparrow
V_b	Decrease	Stay the same	Increase	
V_c	Decrease	Stay the same	Increase (Approx)	
I	Decrease	Stay the same	Increase	



$V_a = \frac{V_e}{2}$ (matched by $\lambda/4$ transformer) $= V_b = 1.75V$

$V_c = -j \frac{V_e \sqrt{2}}{2} = -j \frac{3.5}{\sqrt{2}}$
 $I = 0$



$V_a = \frac{V_b}{2} = -V_b = -1.25 = V_a$
 $1.25 = V_b$

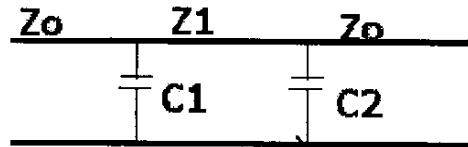
$V_c = 0$

$I = \frac{2V_a}{100 \Omega} = \frac{-2.5}{100} A = -.025 A$

V

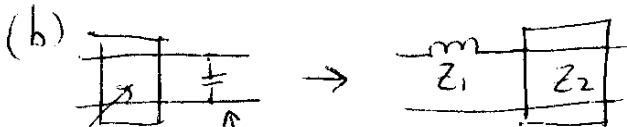
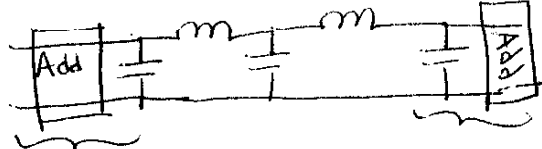
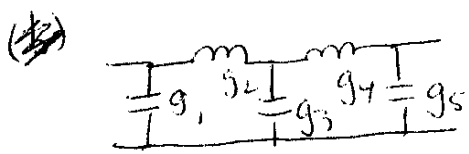
Problem 4 (50 points) Filters

- (a) Find the normalized coefficients ("g" values) for a maximally flat low pass filter that has a cutoff frequency of 1 GHz and passes less than -40 dB at 3 GHz.
- (b) Demonstrate how to apply the Kuroda identities to this filter design. Numerically compute the values of one of the identities, and demonstrate the rest simply by drawing them.
- (c) Show how to implement the system below using open circuited stubs.



(a) From Fig 8.26 p 450 $|\frac{\omega}{\omega_c}| - 1 = |\frac{3}{1}| - 1 = 2.0$

For 40dB, you need a 5th order filter From Table 8.3 p 449
 $g_1 = 0.6180, g_2 = 1.6180, g_3 = 2, g_4 = 1.6180, g_5 = 0.6180, g_6 = 1.0$

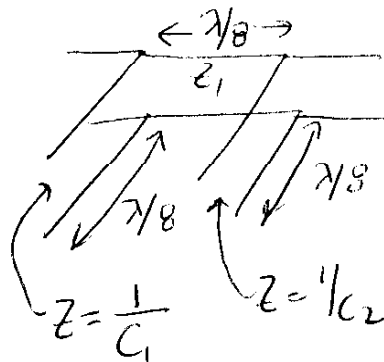
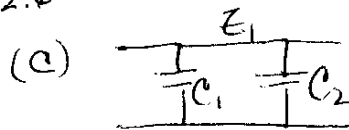
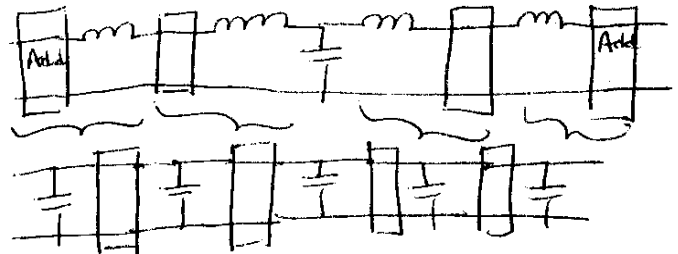


$$n^2 Z_1 = 1 \quad \frac{1}{n^2 Z_2} = g_1$$

$$n^2 = 1 + Z_2 / Z_1 = 1 + \frac{1}{n^2 g_1} = 1 + \frac{1}{g_1} = 2.6$$

$$Z_1 = \frac{1}{n^2} = .382$$

$$Z_2 = \frac{1}{g_1 n^2} = .618$$



Problem 5 (50 points) Waveguides

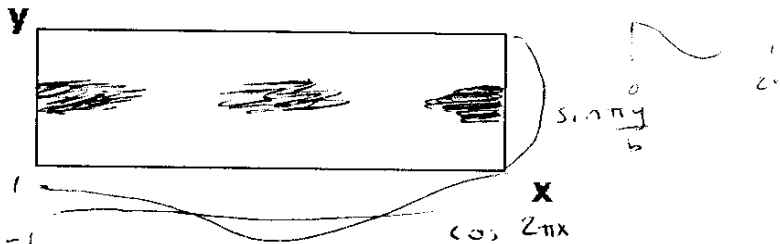
A rectangular waveguide has a cross section of 12 cm x 4 cm and is filled with ceramic that has $\epsilon_r = 9.0$

10 (a) Calculate the cutoff frequencies for the first four modes.

Mode	Fc (Hz)
TE ₁₀	0.41 G
TE ₂₀	0.83 G
TE ₀₁ , TE ₃₀	1.24 G
TE ₁₁ , TM ₁₁	1.30 G

20 (b) Sketch the $|E_x|$ for the TM₂₁ mode by shading the areas with high $|E_x|$ dark and other areas light.

$$E_{x, TM_{21}} = K \cos \frac{2\pi x}{a} \sin \frac{\pi y}{b} e^{-j\beta z}$$



20 (c) Sketch a method of feeding the TM₂₁ mode using aperture coupling for the electric field that DOES NOT allow the TE₂₁ mode to be excited.

$$(a) f_c = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} = .0158 \times 10^9 \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

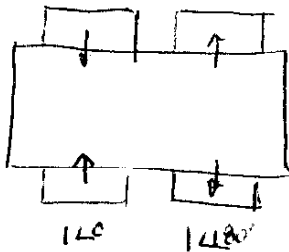
$$\mu = 1.26 \times 10^{-6}$$

$$\epsilon = \epsilon_r \epsilon_0 = (9)(8.854 \times 10^{-12})$$

$$a = 12 \text{ cm} \quad b = 4 \text{ cm}$$

10	.41 G
20	.83 G
30	1.24 G
40	1.65 G
11	1.30 G
01	1.24 G

(6)



For example
(But it allows
TE₂₁ to
propagate)

Problem 6 (50 points) Noise

Find the effective noise temperature of the WLAN receiver discussed in class. Consider one arm of the receiver containing an antenna, amplifier, splitter, filter, detector.

Given:

The ambient temperature is 290 K.

The antenna is a monopole with a gain of 3 (linear, not dB)

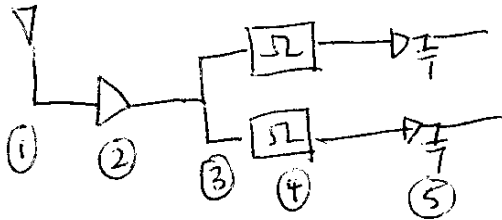
The amplifier is the only active device and has a noise temperature of 350 K. It has a gain of 15 dB.

The power divider is a 3dB splitter.

The filters are coupled line bandpass filters with a gain of -5 dB in the passband.

The diode detectors are perfectly matched (\odot) and have a gain of 0 dB.

$$T_{cas} = T_{e1} + \frac{T_{e2}}{G_1} + \frac{T_{e3}}{G_1 G_2} + \frac{T_{e4}}{G_1 G_2 G_3} + \frac{T_{e5}}{G_1 G_2 G_3 G_4}$$



$$\begin{aligned} T_{e1} &= 290^\circ\text{K} & G_1 &= 3 \\ T_{e2} &= 350^\circ\text{K} & G_2 &= 10^{15/10} = 31.6 \\ T_{e3} &= 290^\circ\text{K} & G_3 &= 10^{-3/10} = 0.5 \\ T_{e4} &= 290^\circ\text{K} & G_4 &= 10^{-5/10} = 0.316 \\ T_{e5} &= 290^\circ\text{K} & G_5 &= 1.0 \end{aligned}$$

$$\begin{aligned} T_{cas} &= (290) + \frac{350}{3} + \frac{290}{(3)(31.6)} + \frac{290}{(3)(31.6)(0.5)} \\ &\quad + \frac{290}{(3)(31.6)(0.5)(1.0)} = \boxed{421.96^\circ\text{K} = T_{cas}} \end{aligned}$$

Name _____

Problem 1 _____ / 50 points

Problem 2 _____ / 50 points

Midterm I replacement total _____

Problem 3 _____ / 50 points

Problem 4 _____ / 50 points

Midterm II replacement total _____

Problem 5 _____ / 50 points

Problem 6 _____ / 50 points

Midterm III replacement total _____